

# HT800" Portable Radios

THEORY/MAINTENANCE MANUAL

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# **FOREWORD**

#### SCOPE OF MANUAL

This manual offers only theory and maintenance information for the equipment listed. Service diagrams, parts lists, and printed circuit board details are not a part of this manual, but are contained in a supplementary service manual as listed on the contents page.

#### **NOMENCLATURE**

Motorola equipment is specifically identified by the model number on the nameplate.

NOTE: Be sure to use the entire model number when making inquiries about your equipment.

Identifiers have been assigned to chassis and kits. Use these identifiers when requesting information or ordering replacements.

#### **PRODUCTION CHANGES**

When production and engineering changes are incorporated into the equipment, a revision number is assigned to the chassis or kit affected; -1, -2, -3, etc.

The chassis number complete with revision number, if any, is stamped on the chassis at the time of production. The revision number becomes an integral part of the chassis identifier. Revisions, if any, are listed on the schematic diagram.

#### **MANUAL REVISIONS**

Changes which occur after a manual is printed are described in the Manual Revision. These "FMRs" give the reader complete information on the change including pertinent parts listing data.

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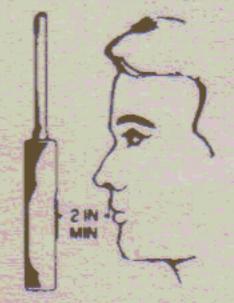
#### **National Service Manager**

Motorola Communications & Electronics, Inc. 1301 Algonquin Road, Schaumburg, IL 60196

#### **SAFETY INFORMATION**

The Federal communication Commission (FCC) with its action in General Docket 79-144, March 13, 1986 has adopted a safety standard for the human exposure to radio frequency (rf) electromagnetic energy emitted by FCC regulated equipment. Motorola subscribes to the same safety standard for use of its products. Proper operation of this radio will result in user exposure substantially below the FCC recommended limits.

DO NOT hold the radio with the antenna very close to, or touching, exposed parts of the body, especially the face or eyes, while transmitting. The radio will perform best if the microphone is two or three inches away from the lips and the radio is vertical.



DO NOT hold the transmit (PTT) switch on when not actually desiring to transmit.

DO NOT allow children to play with any radio equipment containing a transmitter.

DO NOT operate a transmitter near unshielded electrical blasting caps or in an explosive atmosphere unless it is a type especially qualified for such use.

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# HT800™ Portable Radios Handie-Talkie® Portable Radios

#### CONTENTS

FOREWORD	inside front co	ver
LIST OF ILLUSTRATIONS	wood work oo	ii
LIST OF TABLES		ii
VHF MODEL CHART		iii
UHF MODEL CHART		iv
ACCESSORIES		v
SPECIAL TERMS AND ABBREVIATIONS		vi
GENERAL DESCRIPTION		27)
1. INTRODUCTION		1
INTRODUCTION      STANDARD FEATURES		2
3. SPECIAL STANDARD FEATURES	***************************************	2
4. PRINTED CIRCUIT BOARDS AND FLEXIBLE CIRCUITS		2
5. BATTERIES	***************************************	3
BATTERY CHARGING		
1. CHARGERS AVAILABLE		3
2. BATTERY CONSTRUCTION		3
3. BATTERY CHARACTERISTICS		4
4. MAINTENANCE		4
5. STORAGE		
6. DETERMINING BATTERY CAPACITY		4
THEORY OF OPERATION		
1. INTRODUCTION		5
2. BASIC FUNCTIONAL DESCRIPTION		5
3. DETAILED CIRCUIT DESCRIPTION		8
MAINTENANCE		
1. INTRODUCTION		20
2. PREVENTIVE MAINTENANCE		20
3. DISASSEMBLY		20
4. REASSEMBLY		23
5. SAFE HANDLING OF CMOS DEVICES		25
6. REPAIR PROCEDURES AND TECHNIQUES	***************************************	25
7. TEST EQUIPMENT AND SERVICE AIDS		26
8. FIELD PROGRAMMING		27
TROUBLESHOOTING		
1. INTRODUCTION		28
2. TROUBLESHOOTING PROCEDURE	***************************************	28
3. VOLTAGE MEASUREMENT AND SIGNAL TRACING		29
4. TROUBLESHOOTING CHARTS		29
REPLACEMENT PARTS ORDERING	inside back co	ver

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# LIST OF ILLUSTRATIONS

FIGURE	TITLE	AGE
1.	Typical HT800 Series Radio	1
2.	Top Panel Controls and Indicators	1
3.	Typical Battery Construction, Rear View	3
4.	DC Voltage Distribution Block Diagram	5
5.	Frequency Generation Circuits	6
6.	Receiver Block Diagram	6
7.	Transmitter Block Diagram	7
8.	VCO/Synthesizer Block Diagram, VHF	9
9.	VCO/Synthesizer Block Diagram, UHF	10
10.	Microcomputer Interface	11
11.	U1, I-F Module	14
12.	Receiver Audio Circuitry	15
13.	Transmitter Audio Circuitry	16
14.	DTMF Option Diagram	17
15.	Simplified Multicall Circuitry	19
16.	Troubleshooting, Test Equipment, and Programming Setup Detail	26
	LICT OF TABLES	
	LIST OF TABLES	

TABLE	TITLE	PAGE
1.	Batteries for the HT800 Radio	3
2.	Test Equipment	27

# **VHF MODEL CHART**

						JEL	LH	ART		·				POWER OUT	NO. OF CHANNELS	CHANNEL SPACING	
H.	338													2 WATTS	16	20/25KHz	
	H3		NU!	201011										2 WATTS	2	20/25KHz	
		H3	38	-	_									2 WATTS	8	20/25KHz	
-			H3	38	NU	950	OA.	N						2 WATTS	16	12.5kHz	
				H3	38	NU	952	20A	N_				•	2 WATTS	2	12.5KHz	
		H33SNU9580AN H43SNU9100AN												2 WATTS	8	12.5KHz	
			-			H	43S	NU	910	OA	N			5 WATTS	16	20/25kHz	
							H	435	NU	912	OAN			5 WATTS	2	20/25KHz	
								H	138	NU	9180	AN		5 WATTS	8	20/25KHz	
1 24									H	43S	NU9	500AN		5 WATTS	16	12.5kHz	
100	-		- 1							H	43SN	IU9520	DAN	5 WATTS	2	12.5KHz	
									-		H43	SNU9	580AN	5 WATTS	8	12.5KHz	
					0.5								ITEM NO.		DESCRIPTION		
١	Α	Α	Α	Α	Α	Α	Α	Α	Α	Α	Α		NAD6282A	ANTENNA, H	ELIFLEX (136-151MHz CC	DED YELLOW)	
1	Α	Α	Α	Α	Α	Α	Α	Α	A	A	A		NAD6283A	ANTENNA, H	ELIFLEX (146-162MHz CC	DED BLACK)	
1	Α	Α	Α	Α	Α	Α	A	Α	Α	Α	Α		NAD6284A	ANTENNA, HELIFLEX (162-174MHz CODED BLUE)			
(	X	X	Χ	Χ	X	X	X	Х	X	X	X		NHN6419A	HOUSING, SHADOW BRONZE			
(	X	X	X	X	X	X	X	X	X	X	X		NHN4623A	HARDWARE,	MISCELLANEOUS		
	Χ			Χ			Χ			X			NTN4749A	ESCUTCHEO	N, 2-CHANNEL		
(	X	X	Χ	X	X	X	X	X	X	X	Х		NTN4767A	LABEL, FCC			
(	X	X	X	Χ	X				ev.				NTN4822A	BATTERY, RA	PID CHARGE, MEDIUM C	APACITY	
(	Χ	X	X	Χ	Χ	X	X	X	X	X	Х		NTN4904A	FASTENER, CIRCUIT BOARD			
(	Х	Χ	Х	Χ	X	Χ	Χ	Х	X	Х	X		NTN4924B	BELT CLIP, 3-	INCH		
		Χ			X			Х			Х		NTN4928A		N, 8-CHANNEL		
(	Χ	Χ	Х	X	Χ	X	X	Х	Х	X	Х		NTN4956A	FRONT COVE			
	X	Χ	X	Χ	X	X	X	Х	X	Х	X		NTN5096A	NAMEPLATE			
	Х	Χ	v	X	Χ								NTN5101A	FRAME, CON	TROL TOP		
							X	Х		X	Х		NTN5104A	FRAME, CON			
	Х	Χ	Χ	Х	Χ	X	Х	Х	X	X	X		NTN5374A	CONTROLLE			
(	Χ	Χ	Χ	Х	Χ	X	X	X	X	X	X		NTN5394A	SHIELD			
			Х			Χ			X				NTN5430A		TROL TOP (INCLUDES 16-	CHANNEL ESCUTCHEON	
1						Χ	Χ	X	Χ	X	Х		NTN5447A		ID CHARGE, HIGH CAPACIT		
			Α	Α	Α	1							NUD6891A		R, 2-WATT, 12.5KHz	(136-151MHz)	
			Α	Α	Α								NUD6892A		R, 2-WATT, 12.5KHz	(146-162MHz)	
		(jaki)	Α	Α	Α								NUD6893A		R, 2-WATT, 12.5KHz	(157-174MHz)	
									Α	Α	Α		NUD6901A		R, 5 WATTS, 12.5KHz	(136-151MHz)	
1								-	Α	A	Α		NUD6902A		R, 5 WATTS, 12.5KHz	(146-162MHz)	
1									Α	A	Α		NUD6903A		R, 5 WATTS, 12.5KHz	(157-174MHz)	
	Α	Α								1 - 1	17.7	++	NUD6961A		R, 2-WATT, 20/25KHz	(136-151MHz)	
		A								-			NUD6962A		R, 2-WATT, 20/25KHz	(146-162MHz)	
		Α								7 ·			NUD6963A		R, 2-WATT, 20/25KHz	(157-174MHz)	
						Α	Α	Α		-			NUD6971A		R, 5 WATTS, 20/25KHz		
1						A	A	A		7		++	NUD6971A		R, 5 WATTS, 20/25KHz	(136-151MHz)	
_		-	-	-	-	A	A	Α					NUD6972A NUD6973A		R, 5 WATTS, 20/25KHz	(146-162MHz) (157-174MHz)	

KEY X = ITEM INCLUDED

A = ALTERNATE ITEM INCLUDED, CHOICE DEPENDS ON CARRIER FREQUENCY AND TRANSMIT POWER

# **UHF MODEL CHART**

			.:			)EL	CH	ART		<u>.</u>	L			POWER OUT	NO. OF CHANNELS	CHANNEL SPACING
13		-	910											2 WATTS	16	20/25KHz
	НЗ		NU		1000	_	D. F							2 WATTS	. 2	20/25KHz
		H3		_		OAI								2 WATTS	8	20/25KHz
			H3	34S	NU	950	0Al	N						2 WATTS	16	12.5KHz
	5			H3	34S	NU	952	OAI	N				1	2 WATTS	2	12.5KHz
	H34SNU9580AN H44SNU9100AN									N				2 WATTS	8	12.5KHz
	H44SNU9100AN H44SNU9120AN							910	OA	N			4 WATTS	16	20/25KHz	
		H					H	148	NU	912	MAO			4 WATTS	2	20/25KHz
								H4	45	NU	9180	AN		4 WATTS	8	20/25KHz
			_						H	145	NU95	MAOO		4 WATTS	16	12.5KHz
		h i						×		H	14SN	U9520	DAN	4 WATTS	2	12.5KHz
		4									H44	SNU9	580AN	4 WATTS	8	12.5KHz
						To All							ITEM NO.		DESCRIPTION	
	X	X	X	X	X	X	X	X	X	X	X		NAE6350A	ANTENNA, W	/HIP (403-512MHz)	
	X	X	Χ	X	X	Χ	X	X	X	X	X		NHN6419A	HOUSING, SI	HADOW BRONZE	
	X	X	X	X	X	X	X	X	X	X	X		NTN4623A	HARDWARE,	MISCELLANEOUS	STATE OF STA
	Х	Χ			X								NTN4749A	ESCUTCHEC	N, 2-CHANNEL	
	X	X	Χ	X	X	X	X	X	X	X	X		NTN4767A	LABEL, FCC		
	X	X	X	X	X							1	NTN4822A	BATTERY, RA	APID CHARGE, MEDIUM C	APACITY
	Х	X	Χ	X	X	X	X	X	X	X	Х		NTN4904A	FASTENER,	CIRCUIT BOARD	
	Х	X	X	X	X	Χ	X	X	X	X	X		NTN4924B	BELT CLIP, 3-	INCH	
1		X			X			X			X		NTN4928A	ESCUTCHEC	N, 8-CHANNEL	
	X	X	X	X	X	X	X	X	X	Х	X		NTN4956A	FRONT COVI	ER	
	X	Χ	X	X	X	X	X	X	X	X	X	1	NTN5096A	NAMEPLATE		
1	Х	X		X	X							$\Box$	NTN5101A	FRAME, CON		
1		-					X	X		X	X		NTN5104A	FRAME, CON		
	X	X	X	X	X	X	X	X	·X	X	X		NTN5374A	CONTROLLE		
1						X	X	X	X	X	X		NTN5425A	SHIELD		
	X	X	X	X	X								NTN5429A	SHIELD		
			X			X			X				NTN5430A		ITROL TOP (INCLUDES 16-	CHANNEL ESCUTCHEON
1						X	X	X		X	X		NTN5447A		APID CHARGE, HIGH CAP	and the second s
+			A	A	A								NUE6971A		R, 2-WATT, 12.5KHz	(403-433MHz)
1			A	A	A				76				NUE6972A		ER, 2-WATT, 12.5KHz	(438-470MHz)
+			7 10						A	A	A		NUE6981A		R, 4 WATTS, 12.5KHz	(403-433MHz)
+										A			NUE6982A		R, 4 WATTS, 12.5KHz	(438-470MHz)
1	Α	A									71		NUE7101A	+	R, 2-WATT, 20/25KHz	(403-433MHz)
1	A		9.										NUE7102B		R, 2-WATT, 20/25KHz	And the state of t
+	,					Α	Α	Α					NUE7102B			(438-470MHz)
						^		A					NUE7111A NUE7112A	TRANSCEIVE	R, 4 WATTS, 20/25KHz	(403-433MHz) (438-470MHz)

KEY X = ITEM INCLUDED

A = ALTERNATE ITEM INCLUDED, CHOICE DEPENDS ON CARRIER FREQUENCY AND TRANSMIT POWER

# **ACCESSORIES**

Motorola offers a variety of accessories for the HT800 Portable Radio to increase communications efficiency. Many of the accessories available are listed below, but for a complete list, consult your Motorola sales representative.

#### Antennas:

NAD6282A VHF, Heliflex (136-151MHz) YELLOW VHF, Heliflex (151-162MHz) BLACK VHF, Heliflex (162-174MHz) BLUE

NAE6132A S.M.A. Antenna

NAE6231A
UHF, Heliflex (403-433MHz) RED
UHF, Heliflex (438-470MHz) GREEN
UHF, Heliflex (470-512MHz) BLACK
UHF, Flexible Whip (403-512MHz)

#### Batteries:

NTN4822A Medium Capacity, Nickel-Cadmium, Rapid Charge

NTN4823A Medium Capacity, Nickel-Cadmium, Rapid Charge, Factory Mutual Approved

NTN4824A High Capacity, Nickel-Cadmium, Rapid Charge NTN5447A High Capacity, Nickel-Cadmium, Rapid Charge

NTN5448A High Capacity, Nickel-Cadmium, Rapid Charge, Factory Mutual Approved

#### **Battery Chargers:**

NLN7967A Wall mount for multi-unit NLN7968A Rack mount for multi-unit

NTN4633A Single-Unit Desk Top (Rapid Rate, 110-Volt)

NTN4634A Single-Unit Desk Top (Rapid Rate, 220-Volt) with European Plug

NTN4635A Single-Unit Desk Top (Standard Rate, 110-Volt)

NTN4636A Single-Unit Desk Top (Standard Rate, 220-Volt) with European Plug

NTN4666A Compact (110-Volt)

NTN4667A Compact (220-Volt) with European Plug NTN4668A Multi-Unit (Rapid Rate, 110-Volt)

NTN4922A Multi-Unit (Rapid Rate, 220-Volt/240-Volt)

#### Carrying Accessories:

NTN4814B Belt Clip Carry Holder NTN4916B 2-inch Belt Clip NTN4924B 3-inch Belt Clip

NTN5450A Leather Swivel Case with T-Strap and DTMF/Multicall Access

NTN5460A Leather Case with T-Strap

NTN5461A Leather Swivel Case with T-Strap

#### Audio Accessories:

NLN8410A Velcro Patch Pin Attachment
NMN6127C Remote Speaker Microphone
NTN4812A Earphone Jack Adapter
NTN5043A Earpiece and Volume Control

NTN5050A Public Safety Remote Speaker/Microphone (requires NAE6132A and NLN8410A)

NTN5075A Universal Connector (Hirose) Adapter

ZMN6031A 2-Wire Surveillance Accessory

#### Miscellaneous Accessories:

NKN6376A Cloning Cable

NKN6408A Mobile Antenna Connector Cable

NTN5368A Mobile RF Adapter

# SPECIAL TERMS AND ABBREVIATIONS

The construction technology and circuits in the HT800 Portable Radio require the use of the following special terms and abbreviations.

Term:

Description:

**Alert Tones** 

Audible annunciators of radio status

Code Plug

That portion of the software that is coded for the individual user

DTMF

Dual-Tone, Multi-Frequency (phone interconnect signalling)

Logic 1

A voltage level of approximately 5Vdc

Logic 0

A voltage level of approximately 0Vdc

PA

**Power Amplifier** 

PL

Private-Line® (tone coded squelch)

PLL

Phase-Locked Loop

RX

Receive

TX

Transmit \*

Transceiver Board

The printed circuit board containing the functional components

of the receiver and transmitter

VCO

Voltage Controlled Oscillator

# **GENERAL DESCRIPTION**

#### 1. INTRODUCTION

The frequency-synthesized HT800 Handie-Talkie Radio is an advanced design, microcomputer-based transceiver that incorporates the latest technology available in two-way radio communications. All channel frequencies and squelch codes are stored in an electrically erasable programmable read only memory (EEPROM), with all transmit and receive operations controlled by a microcomputer.

The functions provided by the radio are identified by the model and option numbers as illustrated by the model/option charts at the front of this manual. Model and option numbers will be shown on the radio's customer information sheet, which is shipped with each new radio.

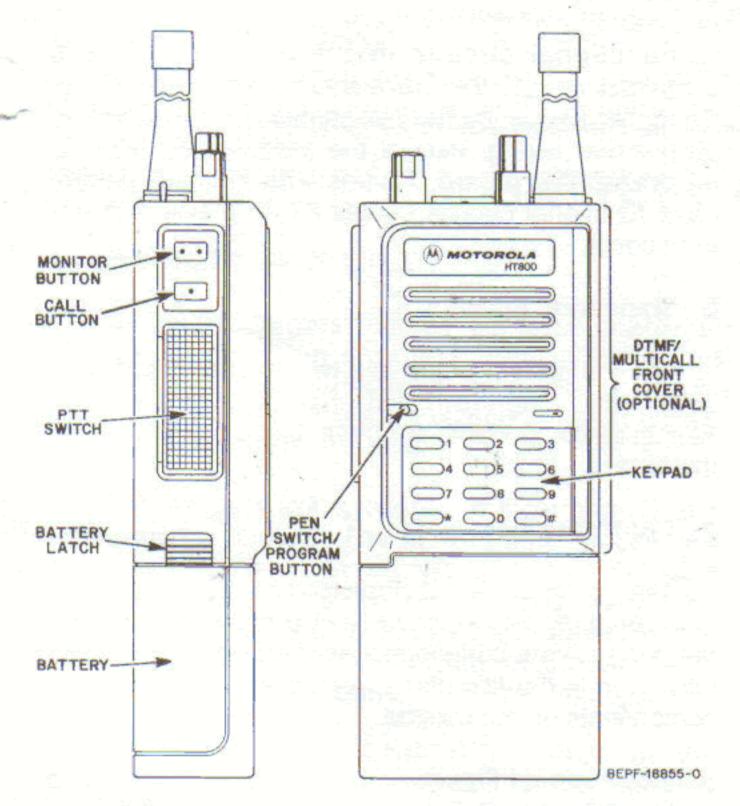
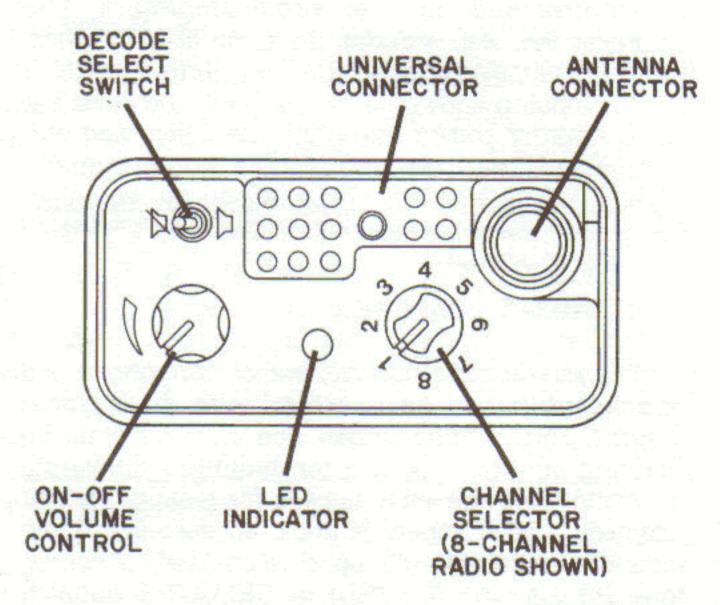


Figure 1. Typical HT800 Portable Radio

## a. Physical Description

All operating controls, except the push-to-talk (PTT) switch, the monitor button, the call button, and the keypad (models with DTMF Option), are located on top of the radio. The PTT switch, monitor button, and call button are located on the left side of the radio (viewed from the front), and the keypad (if so equipped) is an integral part of the front cover.



AEPF-18854-0

Figure 2. Top Panel Controls and Indicators

The HT800 radio is small in size and weight, and constructed of a highly durable, impact resistant, molded polycarbonate housing. O-rings and seals are utilized throughout the radio. All controls, including the PTT switch, the monitor button, the call button, and the keypad are weather resistant, and the microphone and speaker are covered with a special diaphram to provide extra resistance against dirt, dust, and water intrusion. This proven rugged construction offers excellent protection against adverse environmental conditions.

The height of the radio varies with the size of the battery. All other dimensions are standard, except for those radios with a keypad option.

#### b. Electrical Description

Electrically, the radio can be divided into two basic sections; a transceiver board and a controller flexible circuit. The transceiver performs the transmit and receive functions, and the controller controls those functions.

The transceiver board includes an antenna switching circuit, a dual-conversion receiver, and a transmitter. The transmitter carrier and receiver first injection signals are generated by a common phase-locked loop (PLL) consisting of a voltage controlled oscillator (VCO) and a frequency synthesizer.

The controller flex assembly contains a microcomputer, an EEPROM which stores the channel frequencies and squelch codes, and an audio power amplifier integrated circuit (IC) that includes transmitter and receiver audio amplifiers. The controller flex also includes an audio filter IC which encodes and decodes (in conjuction with a microcomputer) squelch codes, adjusts and limits the audio level for correct transmitter deviation, and preemphasizes and de-emphasizes audio signals. Another circuit which is contained on the controller flex is a dc switch, which controls the radio's transmit and receive voltages.

#### 2. STANDARD FEATURES

The HT800 radio has an internal microphone and speaker, but can be operated with an optional external microphone and/or speaker. An external antenna connector and a top-mounted "universal connector" provide easy access for testing, and for attaching a wide variety of audio accessories. Radio models are available with up to 16 channels of carrier, tone Private-Line (CTCSS) or SELECT 5 squelch operation. Type of squelch is enabled on a per channel basis with one code pair available per radio. Two power output levels are offered, medium power (2 watts) or high power (5 watts on VHF models or 4 watts on UHF models).

The battery pack slides on to the bottom of the radio and is held in place by a spring loaded catch. Batteries are available in two different sizes which correspond to the battery capacity (medium and high). The medium and high capacity batteries are available in standard and rapid charge rates. The different size batteries affect the operating time between charges as well as the overall height and weight of the radio.

A bicolor LED on the top of the radio serves as user feedback. The LED indicates when the radio is in transmit (continuous red), a low battery condition (flashing red), or channel busy (flashing green - coded squelch application only).

#### 3. SPECIAL STANDARD FEATURES

#### a. Radio Cloning

Each HT800 radio has a unique data-stored "personality" with frequencies, squelch code pairs, and other operating characteristics. Using a simple cloning cable, one radio's characteristics can be duplicated into another HT800 radio of the same bandsplit.

# b. Field Programming

The HT800 radio utilizes a reprogrammable EEPROM codeplug, which permits operating characteristics to be changed without opening the

radio. Programming is accomplished via a programming cable interface to an IBM PC, Laptop PC, or Personal System/2 computers.

#### SELECT 5 Coded Squelch And Tone "Private-Line" Coded Squelch

Coded squelch allows only those calls with a radio's particular code to be heard, and can be enabled on a per channel basis. So an HT800 radio can have carrier squelch on some channels, SELECT 5 squelch on others, and Tone PL squelch on even others. You can choose from any of the standard European SELECT 5 Signalling formats and 42 Tone Private-Line codes.

# 4. PRINTED CIRCUIT BOARDS AND FLEXIBLE CIRCUITS

#### a. General

Functional circuits in the HT800 radio are contained on: (1) the Transceiver Board and (2) the Controller Flex. Five flexible printed circuits eliminate all discrete wiring, except the switched B+ wire to the transceiver board. Radios with keypad options have functional circuits contained on a board in the front cover.

#### b. Tranceiver Board

The transceiver board is a two-layer printed circuit board containing the rf and i-f portions of the radio. Almost all components are mounted on the top side of this board.

#### c. Controller Flex

The controller flex is packaged inside a protective flex carrier. It is a two-layer flexible printed circuit with the components surface-mounted on one side. When packaged in the flex carrier it is folded in half with all components on the outside.

#### d. Interconnect Flexes

The interconnect flexes are two-layer flexible printed circuits. These include:

- PTT/B+ Flex
- Volume Pot Flex
- Frequency Switch Flex
- I-F interconnect Flex
- · Front Cover Flex

# e. Keypad Board (Optional)

The keypad option board is a four-layer printed circuit board mounted in the radio's front cover. All components are surface mounted on one side of the board.

#### 5. BATTERIES

The rechargeable nickel-cadmium batteries available for the HT800 radio are listed in Table 1. Battery choice is governed by duty cycle, operating time, and maximum height and weight desired.

Table1	<b>Batteries</b>	for the	HTROO	Radio
Table I.	Dattelles	TOT LITE	HIOUU	madio

MODEL	BATTERY	CHARGE	*TYPICAL OF OPE	
NUMBER	CAPACITY	TIME	2-WATT RADIOS	4- & 5-WATT RADIOS
NTN4822A	MEDIUM	1 HR	8 HRS	5 HRS
NTN4823A	MEDIUM	1 HR	8 HRS	5 HRS
NTN4824A	HIGH	1 HR	**13 HRS	8 HRS
NTN4825A	HIGH	1 HR	**13 HRS	8 HRS
NTN4868A	MEDIUM	16 HRS	8 HRS	5 HRS
NTN4869A	HIGH	16 HRS	**13 HRS	8 HRS

<sup>\*</sup> BASED ON A DUTY CYCLE OF 5% TRANSMIT, 5% RECEIVE, AND 90% STANDBY TIME.

# **BATTERY CHARGING**

### 1. CHARGERS AVAILABLE

Available chargers include compact chargers, single-unit desk top chargers, and multiple-unit chargers that may be mounted on a wall or bench. The multiple-unit chargers will charge up to six nickel-cadmium batteries at one time.

The chargers are available in slow charge and rapid charge models. The slow charge models will charge any of the batteries, with or without the radio attached, in 16 hours. The rapid charge models will charge any of the rapid-charge batteries in approximately one hour, and any of the standard-charge batteries at the standard rate (16 hours).

Refer to the ACCESSORIES page at the beginning of this manual for a list of the available battery chargers and their applications. For further information, contact your Motorola sales representative.

# 2. BATTERY CONSTRUCTION (See Figure 3)

The HT800 rapid-charge battery has four charger contacts, two of which receive the charging current. A third contact connects an internal resistor (R<sub>R</sub>) to the charger, automatically setting the charging current output to match the capacity of the battery. The fourth contact connect an internal thermistor to the charger. The thermistor senses battery temperature and automatically controls the charger output to permit maximum charger output without overheating the battery.

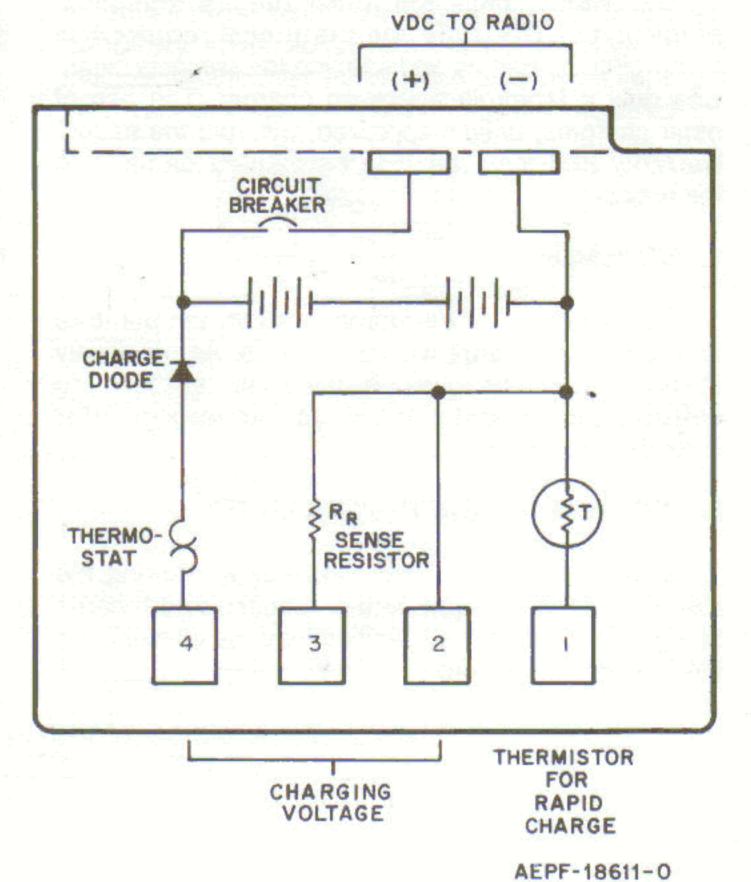


Figure 3. Typical Rapid-Charge Battery Construction, Rear View

<sup>\* 14</sup> HOURS ON VHF RADIOS

All rapid-charge batteries contain an internal current-limiting device (breaker) for protection. A diode in the battery prevents damage from an accidental short between the charging contacts.

#### CAUTION

Sustained shorts across the radio contacts (+, -), excessive current, or excessive heat will destroy the internal thermal fuse, which is not replaceable.

### 3. BATTERY CHARACTERISTICS

Each nickel-cadmium battery consists of eight cells connected in series to provide a nominal 10Vdc output, which remains approximately constant under load until the battery approaches a discharged condition. At this time, a marked decrease in voltage occurs and the discharge condition (1.0 volt per cell) is reached abruptly.

A general characteristic of all rechargeable batteries in storage is self-discharge. If the battery is to be used after an unknown period of storage, it is recommended that it be charged at the full charging rate using an approved battery charger.

#### 4. MAINTENANCE

The battery cells will never require additional electrolyte. The only maintenance required is recharging the battery and keeping the contacts clean. Use only a Motorola approved charger. The use of other chargers, unless approved, will void the battery warranty and may result in permanent damage to the battery.

#### 5. STORAGE

The battery may be stored at room temperature in any state of charge without damage. As previously stated, however, the battery is subject to self-discharge and should be recharged after extended storage.

#### 6. DETERMINING BATTERY CAPACITY

Battery capacity is determined by measuring the time that a fully-charged battery requires to discharge to eight volts through a specified load, as described in the following procedure.

#### NOTE

This procedure requires using a 20-ohm, 1%, 10-Watt load resistor to discharge medium capacity batteries, and an 11-ohm, 1%, 15-Watt load resistor to discharge high capacity batteries.

- a. Obtain a Radio Housing Adapter (Motorola part number 1580368B62) from your nearest Area Parts Office.
- b. Connect the appropriate 20-ohm or 11-ohm load resistor (See Note above) between the gold (+) terminal and a solder lug (-) screw and nut of the housing adapter.
- Connect a voltmeter across the load resistor and slide a fully charged battery onto the housing adapter.
- d. Monitor the voltmeter as the battery discharges through the load resistor, until the voltage is eight volts. This will erase the memory effect.
- e. Disconnect battery from the housing adapter (resistor load) when the cell pack reaches 8.0 volts.

#### CAUTION

Discharging the battery down to 4.0 volts can cause permanent cell pack damage.

- Recharge the battery to a complete charge. This will require a 1-hour rapid charge followed by a 16hour standard charge.
- g. Re-attach the battery to the housing adapter (resistor load) and measure the elapsed time until the cell pack reaches 8.0 volts. Disconnect the battery.
- h. A good battery will require 48 minutes or longer to discharge, indicating greater than 80% of rated capacity. A weak battery will drop below 8.0 volts in less than 48 minutes.

# THEORY OF OPERATION

#### 1. INTRODUCTION

This section of the manual provides a functional description of the HT800 radio. First, overall basic functions are discussed in general terms with each circuit and its relationship to other parts of the radio described. Then, detailed circuit descriptions are given for each board, circuit, and module used in the radio.

#### 2. BASIC FUNCTIONAL DESCRIPTION

# a. DC Voltage Distribution (See Figure 4)

Operating power for the radio is derived from a 10-volt battery. This 10 volts (BATT B+), via the PTT/B+Flex, the Frequency Switch Flex, and the Volume Pot Flex, is applied to the ON/OFF switch. When the radio is turned on, the voltage sources required to operate the various stages of the radio are distributed as shown in Figure 4. In the transmit mode (PTT actuated) a logic low on the R/T line enables the DC switch to provide the required 5 Vdc and 10 Vdc to the transmitter circuits.

# b. Frequency Generation and Distribution Circuits (See Figure 5)

The frequency generation and distribution circuits in the HT800 radio are common to both transmitter and receiver. They consist of two phase-locked loops (PLLs). One PLL provides the carrier frequency for the

transmitter and the injection signal for the receiver first mixer stage. The other PLL generates the second local oscillator (LO) signal. Audio is modulated on the carrier in two different places (two-spot modulation); the VCO's frequency response allows it to modulate audio above 60 Hz, the reference modulator modulates audio below 60 Hz.

The frequency generation circuits include a reference oscillator (U106), a synthesizer (U202), and a VCO (U201). The reference oscillator generates a 16.8 MHz reference signal for the synthesizer. An external adjustment is provided to set the frequency at the output of the reference oscillator.

The following is a functional description of the transmitter first injection PLL. Initially, the VCO becomes active and generates a signal, part of which is coupled back to the synthesizer as a feedback signal. The synthesizer divides this signal and compares it to a reference frequency. If the frequencies differ, the synthesizer generates a control (error) voltage which causes the VCO to change frequency. When the VCO reaches the correct frequency, the synthesizer generates a constant control voltage signal, locking the VCO on frequency. In the transmit mode, voice audio is applied to a varactor on the VCO. The capacitance of the varactor changes in proportion to the instantaneous audio voltage, which results in a shift in carrier frequency at an audio rate. Audio below 60 Hz is modulated onto the synthesizer reference signal, which in turn causes a similar shift in the carrier frequency.

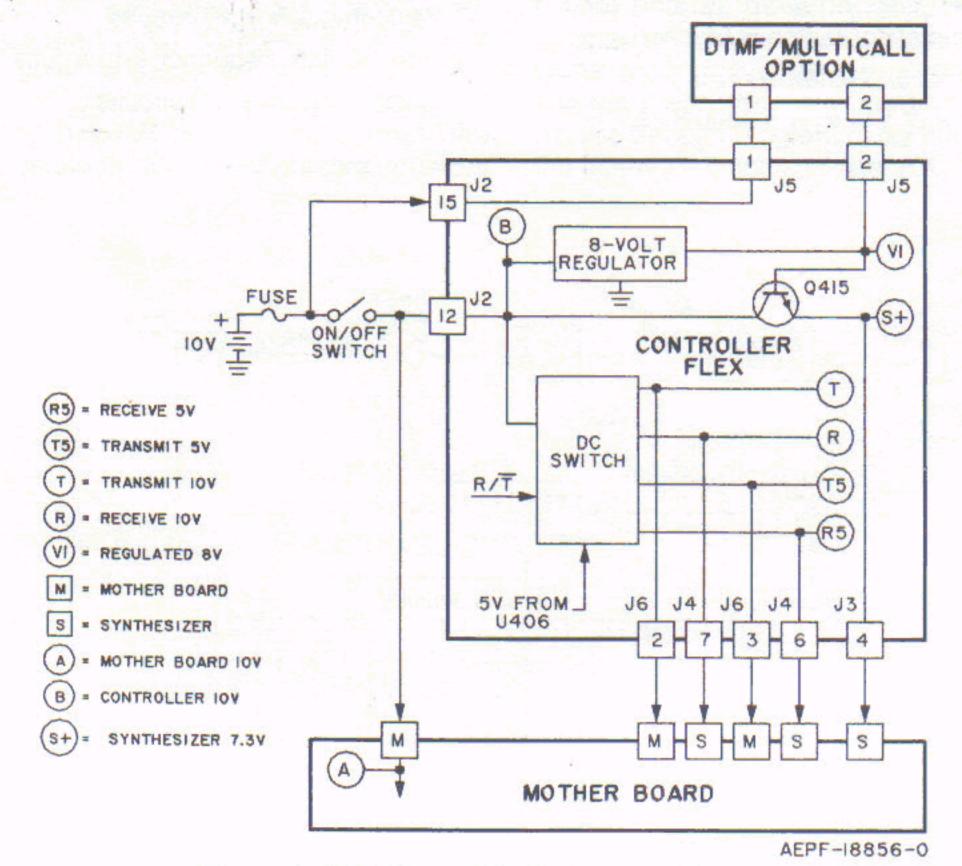


Figure 4. DC Voltage Distribution Block Diagram

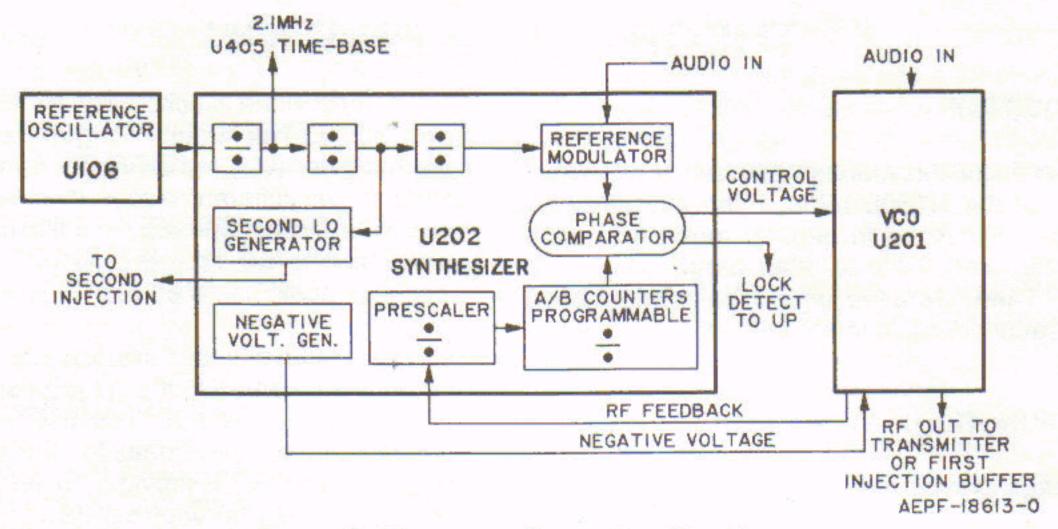


Figure 5. Frequency Generation Circuits

#### c. Basic Controller Functions

Module U401 is a single-chip microcomputer and is the heart of the HT800 controller. It works in conjunction with the 16-channel code plug (U402), which stores radio information data. The controller's functions are as follows:

- Read the PTT and channel selector switches, and program the synthesizer for the desired operating frequency using the information stored in the code plug.
- Set the audio output levels for the VCO and synthesizer.
- · Control the DC switch.
- Unsquelch the receiver's audio PA when a carrier is present, a correct PL tone or SELECT 5 sequence is decoded, or when an alert tone is generated or the monitor button is pressed.
- Monitor the internal and external PTT.
- Encode a PL tone or SELECT 5 sequence.

- · Control the Receive/Transmit LED.
- Monitor battery voltage.
- Perform a self test during power-up.

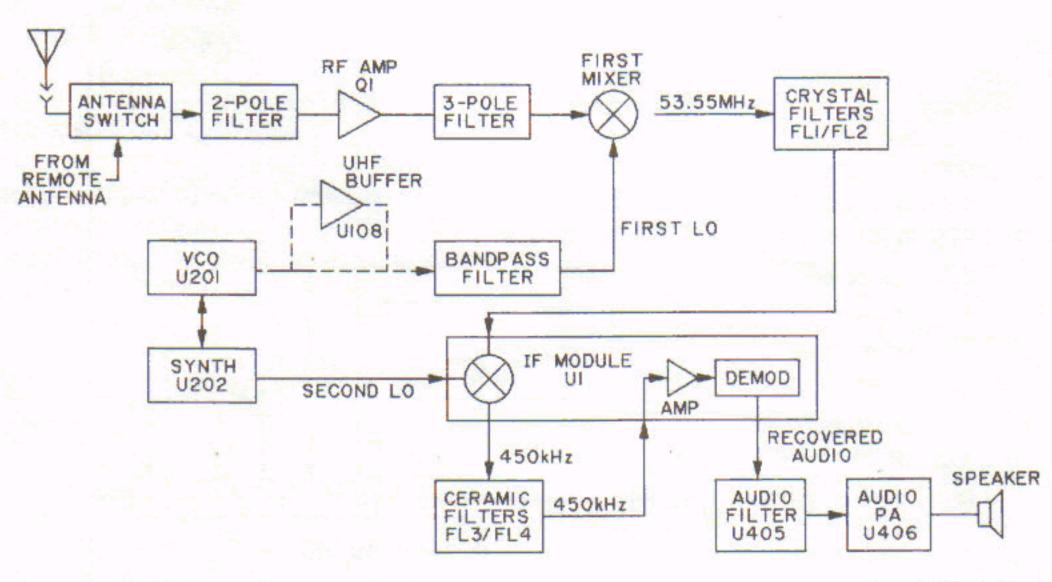
#### d. Antenna Switch

The antenna switch consists of modules U103 and U104 on VHF models and U105 on UHF models. Through the use of pin diodes, the antenna switch directs incoming rf from either the standard or remote antenna to the receiver circuitry and outgoing rf from the transmitter to the remote or standard antenna.

## e. Basic Receiver Operation (See Figure 6)

The HT800 uses double-conversion superheterodyne receiver circuits to provide greater imagesignal suppression and improved adjacent channel selectivity. The receiver consists of three main sections:

- radio frequency (rf) circuits
- intermediate frequency (i-f) circuits
- · audio frequency (af) circuits



AEPF-18614-0

Figure 6. Receiver Block Diagram

#### (1) RF Signal Path

The rf signal is received by the antenna and coupled to a two-pole bandpass filter through the antenna switch. The output of the two-pole filter is amplified by an rf amplifier (Q1). The output of the amplifier is then coupled through a three-pole bandpass filter, and applied to the rf input of the first mixer stage (Q2). An injection signal (FIRST LO) is applied to the second input of the mixer, resulting in an output difference frequency of 53.55 MHz, which is the first i-f frequency.

#### (2) I-F Signal Path

The first i-f signal is passed through highly selective crystal filters (FL1 and FL2) to circuit module U1, where it is mixed with a second oscillator injection signal (SECOND LO) to produce the second i-f frequency of 450 kHz. The low conversion signal is then filtered via highly selective ceramic filters (FL3 and FL4), amplified, and demodulated. The resultant signal (RECOVERED AUDIO) is sent to the audio filter (U405) on the controller flex. Module U1 also contains a squelch detect circuit.

#### (3) Audio Signal Path

Recovered audio from U1 is received by the audio filter IC (U405). The audio filter performs basically two functions in the receive mode. It filters, deemphasizes, and attenuates the voice audio, and routes the signal to the volume control. Secondly, if the radio is receiving a coded signal, U405 low-pass filters the audio and seperates the subaudible PL tones. The tones are filtered, sampled and then sent to the microcomputer for decoding.

After passing through the volume control, the audio is sent to audio PA IC (U406). Integrated circuit

U406 amplifies the audio and drives the speaker. The audio amplifier consists of three separate amplifiers; an internal speaker amplifier, an external speaker amplifier, and a common amplifier. If the internal speaker is selected it is differentially driven by the internal and common amplifiers. If the external speaker is selected it is driven by the external and common amplifiers.

Squelch circuitry resides in the i-f module (U1). Discriminator noise from U1 is sent to U405, where the noise is passed through a programmable attenuator (squelch control) and sent back to U1. The squelch circuits in U1 detect demodulator signal-to-noise ratio and produce a dc logic output (5 volts when carrier is present). This output is read by the microcomputer, which in turn programs the audio filter (U405) to enable the audio power amplifiers on U406.

#### f. Basic Transmitter Operation (See Figure 7)

The transmitter (excluding the frequency generation and distribution circuits described in earlier paragraphs) comprises two main circuits:

- Audio circuitry
- RF power amplifiers

# (1) Audio Signal Path

When the PTT switch is pressed, audio from the microphone is fed to the input of the mic amplifier in U406. The amplified audio is then sent to an audio filter IC (U405). Integrated circuit U405 preemphasizes, limits, and low-pass filters the audio. IC U405 also generates squelch codes, which are summed with the voice audio. The audio is then passed through programmable attenuators and sent to the reference modulator and VCO to be modulated.

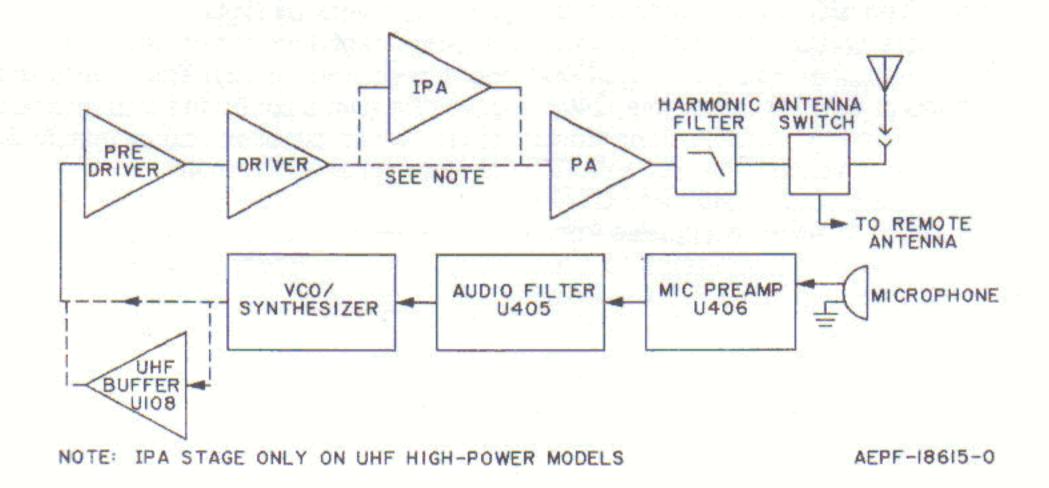


Figure 7. Transmitter Block Diagram

# (2) UHF Medium Power Modulated RF Signal Path

The modulated rf carrier from the VCO / synthesizer is applied through a transmit buffer stage to three consecutive stages of amplification: pre-driver, driver, and rf power amplifier. Medium power UHF radios output 2 Watts of rf power.

# (3) UHF High Power Modulated RF Signal Path

High power UHF radios output 4 Watts of rf power. This is accomplished through four consecutive stages of amplification: pre-driver, driver, intermediate power amplifier (IPA), and final rf power amplifier.

# (4) VHF Modulated RF Signal Path

VHF radios are available in medium and high power models. In both models, the modulated rf carrier is applied directly from the VCO / synthesizer to three consecutive stages of amplification: predriver, driver, and final rf power amplifier. The difference in power output between medium and high power radios is achieved using different final rf power transistors.

#### 3. DETAILED CIRCUIT DESCRIPTION

The circuit descriptions contained in the following paragraphs are supplemented with simplified schematic diagrams to help the service technician understand the signal processing in various parts of the radio. They are not intended for troubleshooting or servicing. Refer to the complete schematic diagram in the service manual when repairing a radio. When signal tracing on the schematic diagram, pay particular attention to the circles and squares around the module's pin numbers. Circles denote connections to the controller flex; squares denote connections to the main circuit board.

#### a. DC Switch

The dc switch controls voltages being applied to the receiver and transmitter circuits. These voltages are R (10V) and R5 (5V) for receive, and T (10V) and T5 (5V) for transmit. The DC switch consists of module U403, transistors Q403 thru Q407, Q412, Q413, Q416, resistors R421 thru R424, R427, R428, and diodes CR403 through CR407. Transistors Q403 and Q405 drive the T voltage line while transistors Q404 and Q406 drive the R voltage line. R5 and T5 voltages are provided via transistors Q407 and Q412 respectively. The DC switch receives its supplies from fused 10V, B, and the five-volt regulator (Q400 collector). Module U403 responds to the R/T line from U405 pin 40, which is controlled by the microcomputer. In transmit the R/T line is low (0 volts) and in receive the R/T line is high (5 volts).

The microcomputer monitors the LOCK DETECT line from the synthesizer (U202 pin 7). When the LOCK DETECT line is low, indicating a frequency lock condition, the microcomputer signals U405, via the microprocessor interface, to switch the output at U405 pin 39 low. This low is applied to diode CR405, which supplies a ground path for the emitters of transistors Q405 and Q406. These emitters must have this ground path so that the R/T line can forward bias Q405 or Q406, activating the T or the R line, respectively.

In transmit (synthesizer locked), the R/T line is at 0 volts. The R/T low is fed to the input of an inverter on U403. The output of the inverter turns on transistors Q416, Q405, and Q403 to activate the T voltage line (Q403 collector). The R/T low is also fed to the base of transistor Q412, which turns on Q412 and activates the T5 voltage line (Q412 collector).

In receive (synthesizer locked, Battery Saver - off, U202 pin 3 high), the R/T line is at 5 volts. The R/T high is fed to the input of U403 where it is NANDed with the high on the BATTERY SAVER line (U403 pins 8 and 7 respectively). The resulting low at the NAND gate output forward biases transistor Q407, which activates the R5 voltage line (Q407 collector). Also, the low output from the NAND gate is inverted and the high output at U403 pin 14 turns on Q406 and Q404 to activate the R voltage line (Q404 collector). When the R voltage line is activated, Q413 is forward biased, which supplies drive voltage for the green LED (CR301A).

If the battery saver option is programmed into the radio, the microcomputer programs the synthesizer to strobe the R and R5 lines via the BATTERY SAVER line. The battery saver signal is a square wave which is NANDed with the 5 volts on the R/T line. The strobing of the receive voltages reduces current drain when the radio is in the stand-by condition.

Another part of the dc switch circuit is an 8-volt regulator. The 8-volt regulator consists of module U407, transistor Q410, and resistors R447, R448, and R446. Module U407 is a five-volt regulator which is offset to 8 volts by R447 and R448. Transistor Q410 is a pass transistor which increases the regulators current sourcing ability. The 8 volts is applied to the audio PA (U406 pin 9) and also applied to the base of Q415, which supplies approximately 6.6V S+ to the synthesizer at U202 pin 4.

# b. Frequency Generation and Distribution (See Figures 8 and 9)

#### (1) The VCO (VHF Radios - Figure 8)

The VCO (U201), in conjunction with the synthesizer (U202) and the reference oscillator (U106), generates rf in both modes of operation (receive and transmit). The VCO RF OUT signal is produced at U201 pin 7. A sample of the rf signal is routed from U201 pin 3 as a buffered feedback to a prescaler circuit in the synthesizer (U202). After frequency comparison in the synthesizer, a resultant control voltage from U202 pin 14 is received at U201 pin 12. This voltage is between 0 and 5 volts when the PLL is locked on frequency. At the same time, a negative voltage from the synthesizer is applied to U201 pin 11. This negative voltage is either -2, -4, -6. or -8 volts. The negative voltage and control voltage are applied at opposing ends of a varactor diode, which tunes the VCO to the correct frequency.

Five volts at U201 pin 8 places the VCO in the receive mode. During the receive condition, the VCO produces the first LO injection signal at U201 pin 7. The signal is routed to the first mixer (Q2), via a transistor buffer stage (Q3).

During the transmit condition, PTT depressed, the five volts at U201 pin 8 is removed and five volts is applied to U201 pin 2. This places the VCO in the transmit mode. During the transmit condition, the VCO generates the carrier signal, and routes it from U201 pin 7 to the pre-driver (Q102), via a transistor buffer stage (Q101). Also in the transmit mode, the audio signal to be modulated onto the carrier is received by a varactor in the VCO module at U201 pin 5.

# (2) The VCO (UHF Radios - Figure 9)

The VCO (U201), in conjunction with the synthesizer (U202) and the reference oscillator (U106), generates rf in both modes of operation (receive and transmit). The VCO rf output, produced at U201 pin 5, is routed to the VCO buffer (U108). A sample of the rf signal is routed from U108 pin 10 (PRE-SCALER RF OUT) as a buffered feedback to a prescaler circuit in the synthesizer (U202). After frequency comparison in the synthesizer, a resultant control voltage from U202 pin 14 is received at U201 pin 13. This voltage is between 0 and 5 volts when the PLL is locked on frequency. At the same time, a negative voltage from the synthesizer is applied to U201 pin 12. This negative voltage is either -2, -4, -6, or -8 volts. The negative voltage and control voltage are applied at opposing ends of a varactor diode, which tunes the VCO to the correct frequency.

In the receive mode, five volts (R5) is applied to U201 pin 8 and U108 pin 3, which places the VCO in the receive mode and enables a receive injection buffer in U108. The rf signal (first LO injection) at U201 pin 5 is received at U108 pin 6 (VCO IN). The buffer stage in U108 routes this signal (RF INJECTION OUT), via U108 pin 4, to the first mixer (Q2).

During the transmit condition (PTT depressed), the five volts at U201 pin 8 and U108 pin 3 is removed. Five volts is applied to U201 pin 3 and U108 pin 9, which places the VCO in the transmit mode and enables a transmit buffer in U108. During the transmit condition, the VCO generates the carrier signal, and routes it from U201 pin 5 to U108 pin 6 (VCO IN). The buffer stage in U108 routes this signal (Tx RF OUT), via U108 pin 8, coil L101, capacitor C105, and coil L116, to the pre-driver Q102. Also in the transmit mode, the audio signal to be modulated onto the carrier is received by a varactor in the VCO module at U201 pin 1.

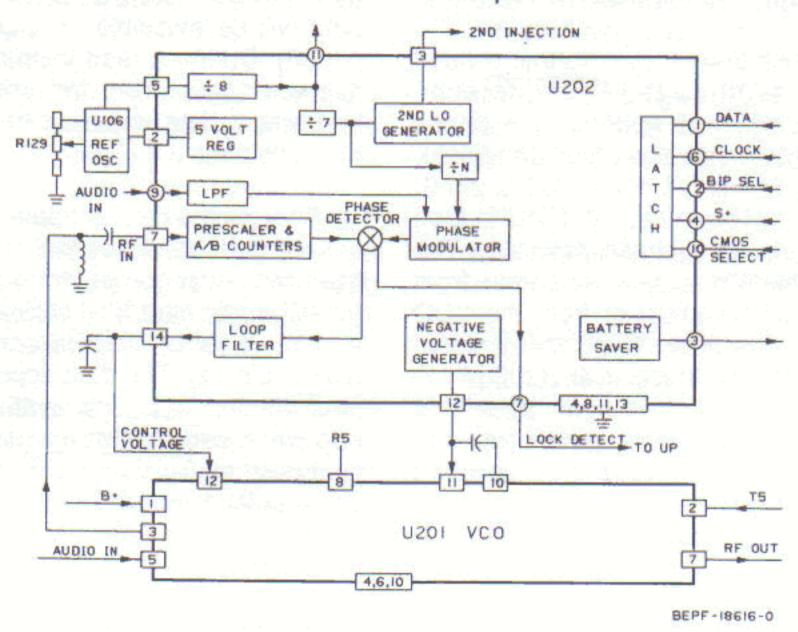


Figure 8. VCO / Synthesizer Block Diagram, VHF

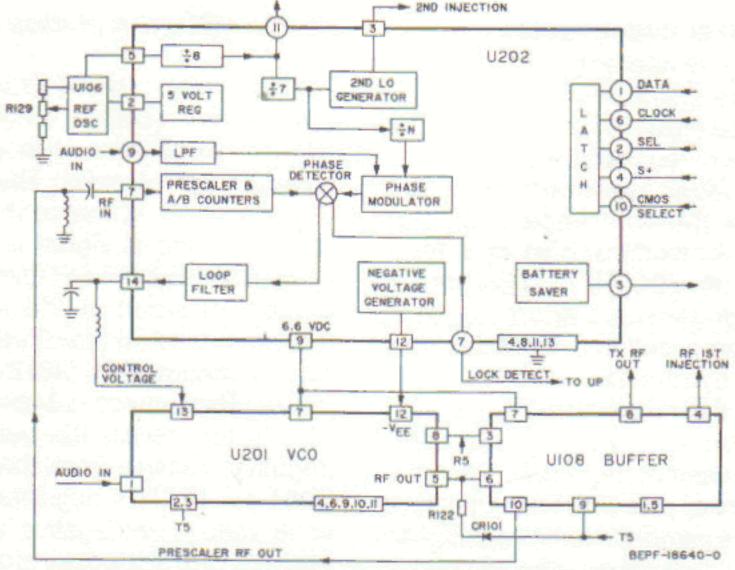


Figure 9. VCO / Synthesizer Block Diagram, UHF

# (3) The Synthesizer

The microcomputer (U401) reads the code plug (U402) and sends set-up signals, which are received by the synthesizer (U202) latch circuit. These set-up signals determine the correct negative voltage and the A/B counter divide ratios needed to generate the proper rf frequencies. The reference frequency for the synthesizer / VCO phase-locked loop is provided by a 16.8 MHz crystal oscillator (U106), which is fine-tuned by resistor R129. The 16.8 MHz crystal oscillator frequency is divided, first to 2.1 MHz and then to 300 KHz. The 300 KHz signal is used for two different applications in the synthesizer.

First, the 300KHz reference frequency is applied to an internal phase-locked loop circuit (within the synthesizer), which generates the receiver's second LO injection signal. The second LO injection frequency of 53.1 MHz (for low-side injection) or 54.0 MHz (for high-side injection) is routed from U202 pin 3 to the second mixer stage in I-F module U1.

Secondly, the 300 KHz frequency is further divided to produce a VCO / synthesizer PLL reference frequency of 5.0 KHz, which is applied to a phase modulator. In the transmit mode, the phase modulator modulates audio below 60 Hz (PL tones, U202 pin 9) onto this reference signal. The reference signal is then fed, as one of two inputs, to a phase detector. The second input signal to the phase detector comes from the VCO (U201 pin 3, VHF radios) or from the VCO buffer (U108 pin 10 UHF radios). This second signal (RF IN) is received by the synthesizer at U202 pin 7, divided by a prescaler circuit, divided again by an A/B counter circuit, and then applied to the phase detector. The phase detector circuit compares the two input signals. If the frequencies are not the same, a CONTROL VOLTAGE (error voltage) is generated and sent to the VCO, ultimately pulling the PLL on frequency. When the two frequencies are the same, the phase detector outputs a low on the lock detect line. This lock detect low is routed to the

microcomputer, which in turn sets up radio transmit and receive voltages. Refer to the "DC SWITCH" section for a more detailed explanation.

#### c. Controller (See Figure 10)

Module U401 is a single chip, 8-bit microcomputer which performs control and processing functions. It works in conjunction with the code plug which stores the radio personality in its non-volatile memory. The microcomputer controls three data buses; the code plug bus, the synthesizer module/audio filter IC bus, and the radio programming bus.

The code plug data bus is bi-directional, meaning that data can be sent to or received from the code plug. When the microcomputer wants to access the code plug it will pull CODE PLUG POWER (U401, pin 39) low, turning on the supply to the code plug through Q402. The microcomputer will then transmit the address of the data to the code plug on the data bus by toggling CODE PLUG CLOCK (U401, pin 34). The data will be available on CODE PLUG DATA (U401, pin 35). During a read instruction, data is input to the microcomputer from the code plug. During a write instruction, data is output from the microcomputer to the code plug.

The synthesizer module/audio filter IC programming bus is uni-directional, meaning that data is sent from the microcomputer to the synthesizer module (U202)/audio filter IC (U405). The bus is synchronous and the flow of data is controlled by SPI CLOCK (U401, pin 14). The data appears on SPI DATA (U401, pins 10 and 11). The synthesizer module has two separate programming latch circuits which are controlled by BIPOLAR SEL (U401, pin 8) and CMOS SEL (U401, pin 7). When programming the synthesizer module, the microcomputer first pulls BIPOLAR SEL low and sends data using SPI CLOCK and SPI DATA. BIPOLAR SEL is then pulled high and CMOS SEL is pulled low. The microcomputer again sends data using SPI CLOCK and SPI DATA.

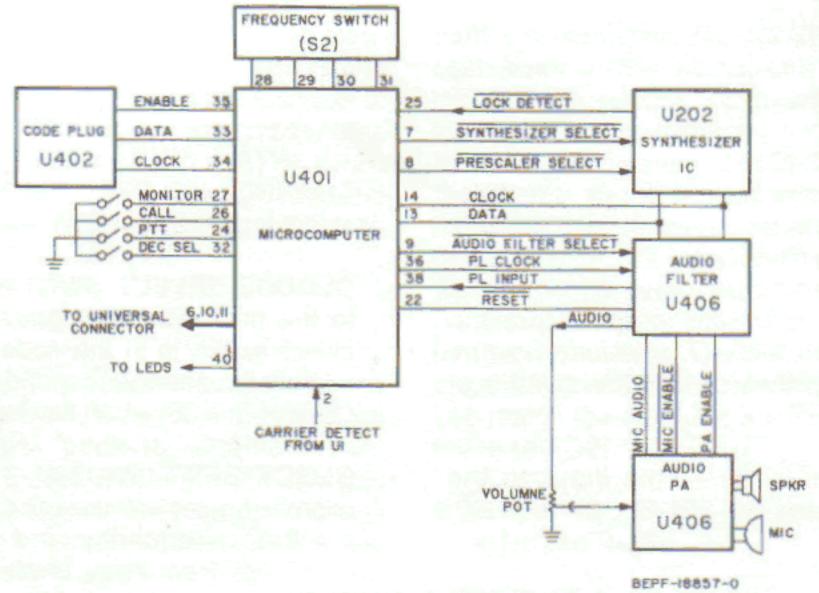


Figure 10. Microcomputer Interface

When the data transfer is complete, CMOS SEL is pulled high. The synthesizer module is now programmed for the new operating frequency. To program the audio filter IC, AF SELECT (U401, pin 9) is pulled low. The data is transfered using SPI CLOCK and SPI DATA. When the data transfer is complete AF SELECT goes high.

The radio programming bus is bi-directional, meaning that data can be sent to or received from the microcomputer. The bus is asynchronous and data is sent or received on SCI DATA (U401, pins 10 and 11). The flow of data is controlled by BUSY (U401, pin 6). A low on the BUSY line indicates that a message exists on the DATA line.

## (1) Microcomputer (U401) functions

- Read the PTT and channel selector status, and program the synthesizer module (U202) for the desired operating frequency using the data stored in the code plug.
- Program the audio filter IC (U405) to set the audio output levels to the speaker and VCO module (U201) and synthesizer module (U202).
- Control the dc switch circuits which supply B+ and other voltages to the receiver and transmitter at various times. It does this by signalling the audio filter IC (U405) to set the R/T line (U405, pin 40) and the dc switch enable line at U405, pin 39.
- Program the audio filter IC (U405) to unsquelch the radio when a carrier is detected, when a squelch code is detected, when an alert tone is to be generated, or when the monitor button is pressed.
- Control the flashing of the LED by turning transistor Q401 on and off.

- (2) Microcomputer (U401) input/output pin functions
- Vss (pin 1) Ground for the microcomputer.
- CARRIER DETECT (pin 2) This input to the microcomputer goes high when a carrier is present. It is used in conjunction with CHANNEL ACTIVITY to determine if the radio should be unsquelched.
- SELECT 5 ENCODE (pins 3, 4) These output lines from the microcomputer encode single tone or SELECT 5 sequences in the form of a tri-state digital waveform.
- SELECT 5 DECODE (pin 5) This input to the microcomputer is a filtered and limited signal from the demodulator used by the microcomputer to decode a SELECT 5 sequence.
- BUSY (pin 6) This line is bi-directional and is used to indicate the presence of data on the programming bus.
- CMOS SEL (pin 7) This output from the microcomputer is used when programming the synthesizer module (U202).
- BIPOLAR SEL (pin 8) This output from the microcomputer is used when programming the synthesizer module (U202).
- AF SELECT (pin 9) This output from the microcomputer is used when programming the audio filter IC (U405) It is also used to reset a watchdog timer in the audio filter IC (U405), ensuring that the microcomputer is operating properly. When the microcomputer is operating properly this line will be pulsed at a periodic rate.

- SCI DATA (pins 10,11) These lines are the asynchronous, bi-directional lines used for communicating with the microcomputer.
- SPI DATA (pins 12,13) These lines are the synchronous uni-directional lines used for communicating with the synthesizer module (U202) and the audio filter IC (U405).
- SPI CLOCK (pin 14) This output from the microcomputer is the clock line used when programming the synthesizer module (U202) or audio filter IC (U405).
- SLAVE SELECT (pin 15) This input to the microcomputer enables the SPI CLOCK and SPI DATA lines.
- ADAPT (pin 16) This output from the microcomputer will go high whenever the channel changes and when going from transmit to receive mode. It will cause the squelch circuitry to go into a fast mode of operation.
- LOW BATTERY (pin 17) This input to the microcomputer goes low when the radio battery voltage drops below approximately 8.5 volts. The microcomputer responds by flashing the red LED when in the transmit mode.
- Pin 18 is not used.
- OSC1, OSC2 (pins 19,20) These two lines are connected to the 3.6864 MHz crystal that provides the reference clock frequency for the microcomputer.
- Vcc (pin 21) 5-volt dc power for the microcomputer.
- RESET (pin 22) A low on this line will reset the microcomputer. The microcomputer is reset by the watch dog timer on the audio filter IC (U404).
- IRQ (pin 23) This pin is not used and is pulled to 5 volts through a resistor.
- PTT (pin 24) This input to the microcomputer goes low when the PTT switch is pressed, and signals the microcomputer to enable the transmitter circuitry.
- LOCK DETECT (pin 25) This input to the microcomputer goes low when the synthesizer is locked on frequency.
- CALL (pin 26) This input to the microcomputer goes low when the call button is depressed. The microcomputer will respond by encoding a call sequence if enabled for the channel.

- MONITOR (pin 27) This input to the microcomputer goes low when the monitor button is pressed. The microcomputer will respond by turning on the audio.
- CHANNEL SELECT (pins 28,29,30,31) Channel selection is made via the freq sw (S3).
- DECODE SELECT SWITCH (pin 32) This input to the microcomputer goes low when the mode select switch is in the coded squelch mode. The microcomputer will respond by turning on the PL CLOCK (pin 36) when carrier is detected.
- CLOCK SHIFT (pin 33) This output from the microcomputer is 0 volts to shift the microcomputer oscillator frequency and 5 volts to keep the oscillator frequency unshifted. The oscillator frequency is shifted depending on the receive frequency of each channel.
- CODE PLUG CLOCK (pin 34) This output from the microcomputer is used to clock data in and out of the code plug.
- CODE PLUG DATA (pin 35) This input/output from the microcomputer receives data from or sends data to the code plug(s).
- PL CLOCK (pin 36) This output from the microcomputer is the reference clock used when encoding/decoding PL.
- CODE PLUG POWER (pin 37) This output from the microcomputer is used to power-up the code plug(s).
- PL DECODE (pin 38) This input to the microcomputer receives filtered and limited squelch code signal from the audio filter IC (U405).
- Pin 39 is not used.
- LED CONTROL (pin 40) This output from the microcomputer turns on the LEDs through Q401.

#### d. Antenna Switch and Filters

#### (1) VHF Radios

The antenna switching circuitry consists of two modules, U103 and U104. Module U103 is the receiver / transmitter signal select switch. Module U104 is the remote / standard antenna select switch. Applying 10V through L116 to U103 pin 1 puts U103 in transmit mode, and creates a low impedance path between pin 1 and pin 2. Removing 10 volts from L116 causes U103 to revert back to receive mode and a low impedance path exists between pin 2 and pin 4.

Grounding pin 3 of U104 selects the remote antenna while an open circuit at pin 3 selects the standard antenna. Ten volts is present at the anode of CR101 during the transmit mode to increase the bias and reduce insertion loss. Coils L115, L119, L120 and capacitors C123, C145, C148, C149, C151, and C152 provide additional filtering and matching to the antennas.

# (2) UHF Radios

In transmit, 10 volts T is supplied to the antenna switch (U105 pin 4), via L114. When T is removed the antenna switch reverts back to receive mode. Grounding the REMOTE ANTENNA SELECT line (pin 7) selects the remote antenna while an open circuit will select the standard antenna. In transmit, with the remote antenna selected, a low impedance path exists between pin 4 and pin 5. When the standard antenna is selected a low impedance path exists between pin 4 and pin 6. In receive, with the remote antenna selected, a low impedance path exists between pin 5 and pin 1. When the standard antenna is selected a low impedance path exists between pin 6 and pin 1. Coils L115 and L122 and capacitors C143 and C154 match the output of U108 to the standard antenna. Capacitors C151 and C185 match the remote port of U108 to the universal connector. When the remote antenna is selected, current flows via R128 and L119 to turn on the remote port. Also, when transmitting with a remote antenna, additional current is provided to the antenna switch via CR103 and R123.

# e. Receiver Selectivity and RF Amplifier

The received signal at the antenna is routed through the antenna switch and antenna matching networks, and applied to the receiver rf front end for filtering and amplification.

#### (1) VHF Radios

There are 5 poles of filtering for rf front end selectivity. Coils L1, L2, and capacitors C1 thru C5 form a two-pole tuned butterworth filter with a bandwidth of greater than 16 MHz. Capacitor C8 thru C14 together with coils L3, L4, and L5 form a 3-pole Chebychev filter with a bandwidth of 16 MHz. The rf amplifier (Q1) is a low noise rf transistor, configured in the common-base mode for good intermodulation performance. Transistor Q1 is biased when the R5 voltage is applied to the resistor divider of R1 and R2. Capacitor C21 provides a good rf ground to the base of Q1. The weak rf signal from the two-pole filter is fed to the emitter of Q1, and the amplified signal is available at the collector.

#### (2) UHF Radios

Tunable preselectors L1 and L2 form a two-pole tunable butterworth filter with a bandwidth of greater

than 8 MHz. Capacitors C1, C2, C3, resistor R1, and coil L3 match the output of the preselector's to the input of the rf amp (Q1). Capacitors C41 and C49 improve the preselector's performance. Transistor Q1 is configured in the common-emitter mode. The amplified rf signal is available at the collector and is matched to the 3-pole fixed tuned preselector (L5, L6, and L7) by L4, C6, and C7. Capacitors C35, C36, and C37 improve preselector performance. In some bandsplits C7, C35, C36, C37, and C41 are replaced by 0-ohm resistors. The 3-pole filter has a bandwidth of greater than 30 MHz. Capacitor C8 and coils L8 and L16 match the output of the 3-pole filter to the input of the mixer (Q2).

## f. Receiver First Mixer, Crystal Filter, and Injection Buffer

#### (1) VHF Radios

Transistor Q2, a dual-gate MOSFET, is used as the first mixer stage. The rf signal from the three-pole filter is fed to the source of Q2. The first injection signal from the VCO, via buffer transistor Q3, is introduced at gate 1. The output of Q2 is taken from the drain. The difference signal of 53.55 MHz is the desired i-f output.

The first or high i-f is fed to filter FL1 / FL2, which is a four-pole quartz crystal filter resonant at 53.55 MHz. The filter provides about 28 dB of adjacent channel protection. Components C20, C19, L12, C37, C24, L7, L8 C25, C26, and C27 provide matching for the crystal filters. The i-f signal is then passed to the i-f module (U1) for further signal processing.

Transistor Q3 is in cascade with an open-collector transistor located within the VCO module (U201). Biasing of Q3 (common base) occurs when the R5 voltage is applied to the voltage divider of R13 and R14. Capacitor C36 insures a good rf ground at the base. Transistor Q3, together with coils L9, L10 and capacitors C30, C32, and C35, provide buffering and rejection of unwanted harmonics on the injection string.

#### (2) UHF Radios

Transistor Q2, a dual-gate MOSFET, is used as the first mixer stage. The rf signal from the three-pole filter is fed to gate 1 of Q2. The first injection signal is developed by the VCO (U201) and sent to an injection buffer contained on the VCO buffer module, U108. The buffered signal is routed through a bandpass filter network consisting of C21, C22, L12, L13, C25, C31 and C30, and applied to gate 2 of the mixer. The output of Q2 is taken from the drain. The difference signal of 53.55 MHz is the desired i-f output.

The first or high i-f is fed to filter FL1 / FL2, which is a four-pole quartz crystal filter resonant at 53.55 MHz. The filter provides about 28 dB of adjacent channel protection. Components L9, L10, L11, L14,

C14, C15, C43, C29, C16, and C18 match the output of the mixer to the input of the i-f module, U1.

# g. Receiver Second I-F and Signal Processing (See Figure 11)

Module U1 contains the second mixer, i-f amplifier, PLL demodulator, noise amplifiers and filters, and squelch circuitry. The first i-f signal (53.55 MHz) is received at U1 pin 7. The second LO injection signal from the synthesizer (U202 pin 3) is received by the mixer at U1 pin 9. The desired output frequency from the mixer is 450 KHz. Therefore, the oscillator injection frequency must be 450 KHz above or below the first i-f of 53.55 MHz. The second oscillator frequency is 54 MHz (high-side injection) or 53.1 MHz (low-side injection). The resulting 450 KHz second i-f signal is filtered by the ceramic filter FL3 and FL4 to reject unwanted mixing products. The second i-f signal is then amplified and can be monitored at M1 (U1 pin 4 or 12). The signal is then demodulated and the resultant audio can be monitored at U1 pin 1. The audio is then passed to the audio filter IC (U405).

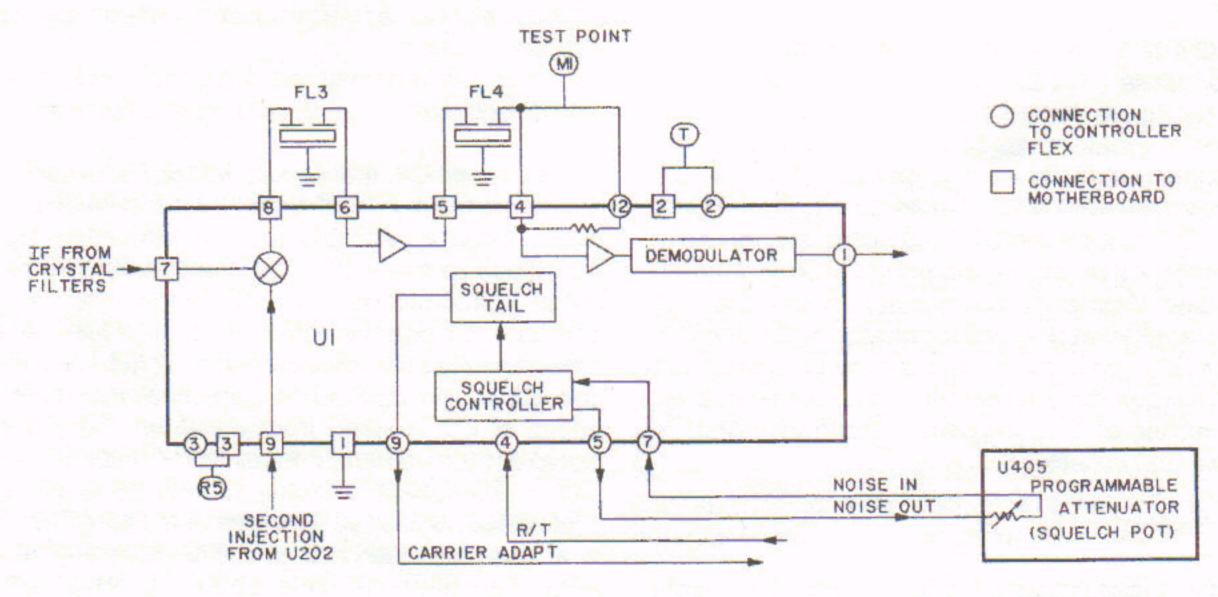
The squelch controller circuit contained in module U1 is a noise detection circuit. The noise output from the squelch controller at U1 pin 5 is routed to U405 to be attenuated by a programmable squelch attenuator and is then fed back to pin U1 pin 7 to the carrier detect circuitry. When the noise level exceeds the threshold level set by the squelch pot on U405, U1 pin 9 (CARRIER DETECT line) goes low, indicating the absence of a carrier signal. The microcomputer reads this CARRIER DETECT low and programs the audio filter (U405) to turn off the power amplifiers on U406 by pulling the PA EN line (U405 pin 3) low. If the noise is less than the threshold level set by the attenuator on U405, U1 pin 9 (CARRIER DETECT line) goes high, indicating the presence of a carrier signal. The microcomputer reads this CARRIER DETECT high and programs the audio filter (U405) to turn on the power amplifiers (U406) by outputting a high PA EN signal (U405 pin 3).

# h. Receiver Audio Circuitry (See Figure 12)

The recovered audio from U1 is routed to the audio filter IC (U405 pins 7 and 8). The audio is low-pass filtered to separate squelch codes and high pass filtered to separate voice. Squelch codes are filtered, sampled, and sent to the microcomputer (U401 pin 38). If the radio is in the PL squelch mode, the microcomputer turns on its decoding circuitry. When the squelch codes are decoded, the microcomputer sends program signals to a microprocessor interface circuit in the audio filter module (U405). The audio filter IC, via the PA EN line, turns on the audio PA IC (U406).

In radios with SELECT 5 decode, the SELECT 5 code sequence is processed by module U801 and sent to the microcomputer (U401, pin 5). Once the proper code sequence is decoded by the microcomputer, program signals are sent from the microcomputer to the microprocessor interface circuit in the audio filter IC (U405) for recognition of either an individual call or a group call. The audio filter IC, via the PA EN line, turns on the audio PA IC (U406). The audio filter IC also outputs the specific call alert tone for individual call or group call, and keys the LED to flash at the proper rate for individual call or group call.

After high-pass filtering, voice audio is deemphasized, filtered, sent through a programmable attenuator (volume control), and then passed from the audio filter to the volume pot (U405 pin 23 to R140). Audio is routed from the volume pot to the audio PA IC (U406 pin 10) and applied to three audio power amplifiers: internal PA, external PA and common PA. The common PA is active for both internal and external speaker applications. Without an external speaker connected, a high input at on the EXTERNAL



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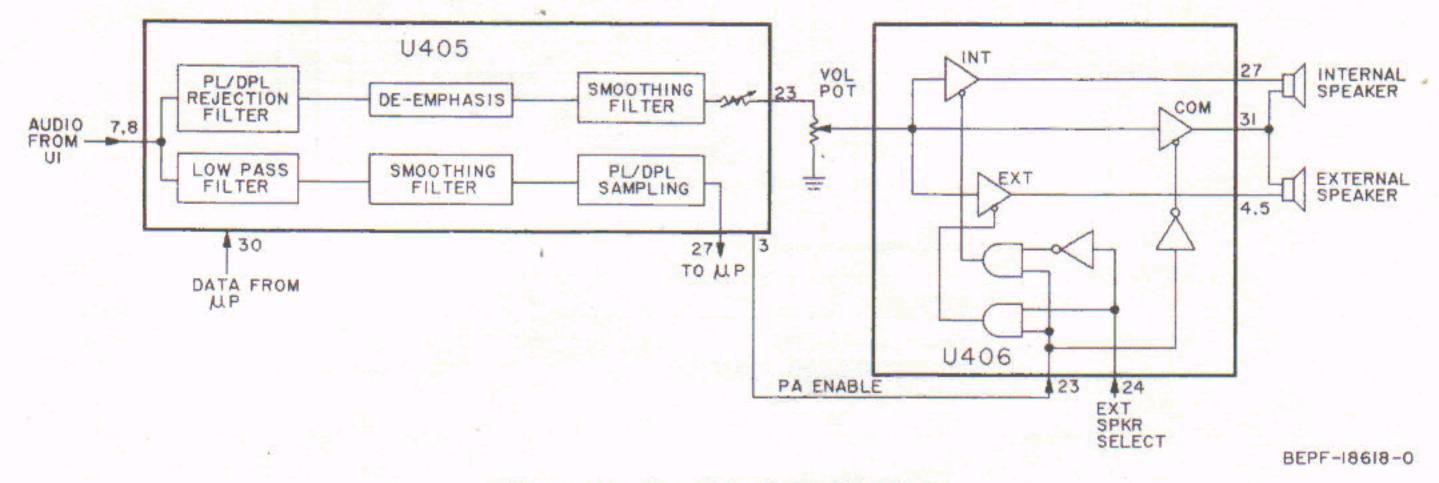


Figure 12. Receiver Audio Circuitry

SPEAKER SELECT line (U406 pin 24) biases the internal PA, and audio from the internal and common power amplifiers is 180 degrees out of phase, which drives the internal speaker (LS1) differentially. Audio from the common power amplifier and external power amplifier is in phase.

If an external speaker is attached to the radio's universal connector, the EXTERNAL SPEAKER SELECT line (U406 pin 24) is pulled low. This low biases the external PA and shifts the audio output of the common amplifier 180 degrees. This phase shift does two things. First, it puts the audio output from the common amplifier 180 degrees out of phase with the audio output from the external amplifier, and the external speaker is driven differentially. Secondly, audio from the common power amplifier and internal power amplifier is in phase, which results in no audio drive for the internal speaker.

# i. Transmitter Audio Circuitry (See Figure 13)

Audio from the microphone is routed to the audio power amplifier (U406), which contains two microphone amplifiers (internal and external). Pressing the PTT switch (internal or external) pulls U401 pin 24 low. The microcomputer reacts by programming the microprocessor interface on U405 to output a low on the R/T line (U405 pin 39). This low is inverted by U403 and applied to U406 pin 18, which enables the microphone circuits. If the internal PTT switch is pressed, a high is present at U406 pin 20, enabling the internal amplifier. If the external PTT switch is pressed, U406 pin 20 is pulled low, and the external microphone amplifier is enabled. Module U406 amplifies and high pass filters the audio. The audio signal is then routed from U406 pin 19 to the audio filter (U405 pin 10), where it is pre-emphasized, limited, and sent through a splatter filter. In PL applications, the audio is summed with the squelch code, which is generated in U405. The audio is then attenuated by two programmable attenuators and the

resultant audio signal is routed from U405 pin 20 to the VCO modulation port (U202 pin 8), and from U405 pin 19 to the reference modulator input at U202 pin 9.

The SELECT 5 encoded sequence consists of one to five tones, defined by the various signalling formats. The encoded sequence is initiated from either the call button or the PTT switch. Depressing the call button results in only the encoded sequence being transmitted with or without voice. If transmitted with voice, the encoded sequence can be tied to pressing the PTT switch (before voice) or releasing the PTT switch (after voice).

The microcomputer encodes a digital waveform and sends the encoded signal, via U401 pins 3 and 4, to shaping and filtering circuits in U801 and the audio filter IC (U405). The signal is attenuated by a programmable attenuator in U405 and the resultant output signal at U405 pin 19 is routed through the synthesizer (U202, pin 8 to pin 1) and applied to the VCO modulation port (U201, pin 1).

#### j. Transmitter

#### (1) VHF Radios

Transmit rf is originated in the VCO / synthesizer modules as discussed in earlier paragraphs of this manual. The rf output of the VCO (U201 pin 7) is applied to the buffer stage (Q101 and associated circuitry). Transistor Q101 is base biased by the T5 voltage via R100, and collector biased by the A voltage line via R102. The rf signal is coupled to the predriver stage (Q102) through C100 and the matching network of C101, L102, and C102. The predriver (Q102), operated class AB, is biased from the T voltage line. Base bias for Q102 is supplied through resistors R105, R104, and R103. Collector bias for Q102 is provided through coil L103. The predriver (Q102) is matched to the driver (Q103) by coils L104, L105, and capacitors C140, C107, C108

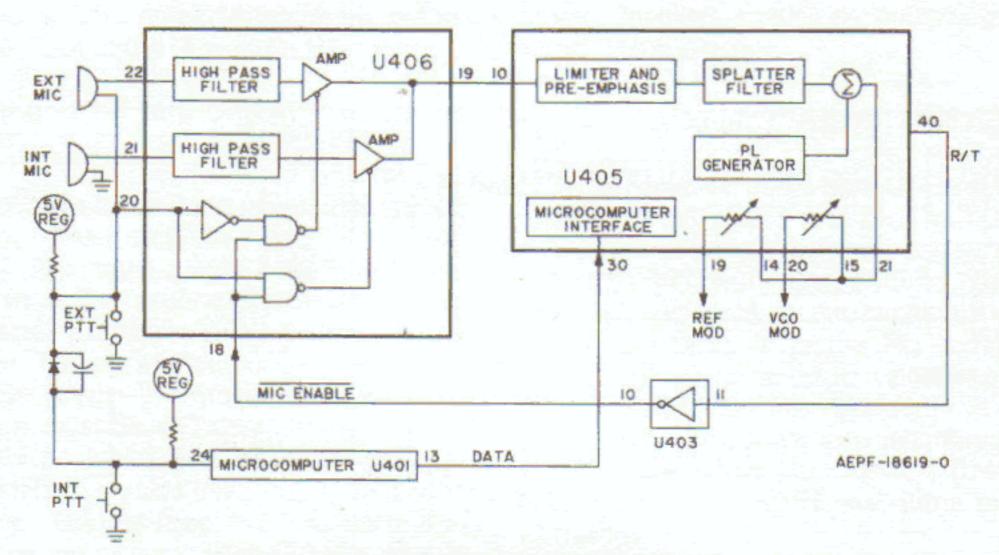


Figure 13. Transmitter Audio Circuitry

and C109. Transistor Q103 operates class C and is supplied from the A voltage line. Collector bias for Q103 is provided through L107. Coil L106 and resistor R108 establish a dc ground return for the base. RF from the collector of Q103 is coupled through C111 and matched to the final PA (Q104) by capacitors C142, C112 and coil L108. Resistor R112 provides stage stability. Transistor Q104 operates class C and is supplied from the A voltage line. Collector bias for Q104 is provided through coil L110. Resistor R109 and coil L109 provide a dc ground return for the base. The PA (Q104) is matched to the antenna switch (U103) by capacitors C143, C117, C118, C120, and coils L111 and L112. A five-element low-pass filter is used to reject unwanted harmonics of the carrier signal. This network consists of C119, C121, C122, L113, and L114. The trimmer cap (C120) is used for adjusting output power. The VHF transmitter is available in 2-watt and 5-watt versions. The main difference is the device used for the final PA (Q104).

#### (2) UHF Low-Power Radios

Transmit rf is originated in the VCO / synthesizer / VCO buffer stages as discussed in earlier paragraphs of this manual. RF output of the VCO buffer (U108 pin 8) is applied to the predriver stage (Q102). Impedance matching between the VCO buffer (U108) and the predriver (Q102) is accomplished by L101, C105, C102 and L116. Transistor Q102 operates in the class AB mode. Transistor Q106 and associated circuitry is use to supply operating voltage to the predriver. The T voltage line forward biases Q106 and supplies drive to the base of Q102 through resistors R107, R103, and R102. The collector of Q102 is biased via L104. Both the buffer and predriver have a 30 MHz bandwidth. The output of the predriver (Q102) is matched to the input of the driver (Q103) by L105, C110, C158 and L106. Capacitor C152 and resistor R121 provide stage stability. Both the driver and final PA (U102) operate in class C mode. The driver (Q103) is collector biased from the A voltage line through coil

L108. A dc ground return for the base is provided by L107 and R113. Resistor R127 and capacitor C155 provide stage stability. RF output from the driver (Q103) is matched to the input of the PA module (U102) by C116, L112, and C117. The trimmer cap on U102 adjusts the power output. The nominal power output for the UHF low power radio is 2 watts.

#### (3) UHF High-Power Radios

The transmitter in high power radios is very similar to the transmitter in low power radios. The differences are:

- In the high power model Q106 is eliminated and the predriver is powered directly from the T voltage line.
- The high power model has an additional stage of amplification (the IPA stage). Impedance matching the output of the driver (Q103) to the input of the IPA (Q104) is accomplished by L118, C167, C111, L109, and C112. Transistor Q104 operates in the class C mode and is supplied by the A voltage line. Collector bias is provided through L111, and a dc ground return for the base is provided through L110 and R115. The nominal output power of the UHF transmitter is 4 watts.

# k. Dual-Tone Multiple Frequency (DTMF) Circuits (Optional)

#### (1) Timed Tone Option

The DTMF circuit receives its power from unswitched battery B+ and an 8-volt regulator (U407) on the controller flex, via connector plug P701. When the radio is turned on, the regulated 8 volts supplied to the DTMF board is routed through a low-pass filter network (R729 and C725). The 8 volts is applied to audio amplifier U702 pin 6, and to the 5-volt regulator (U706). The regulated 5 volts is used throughout the

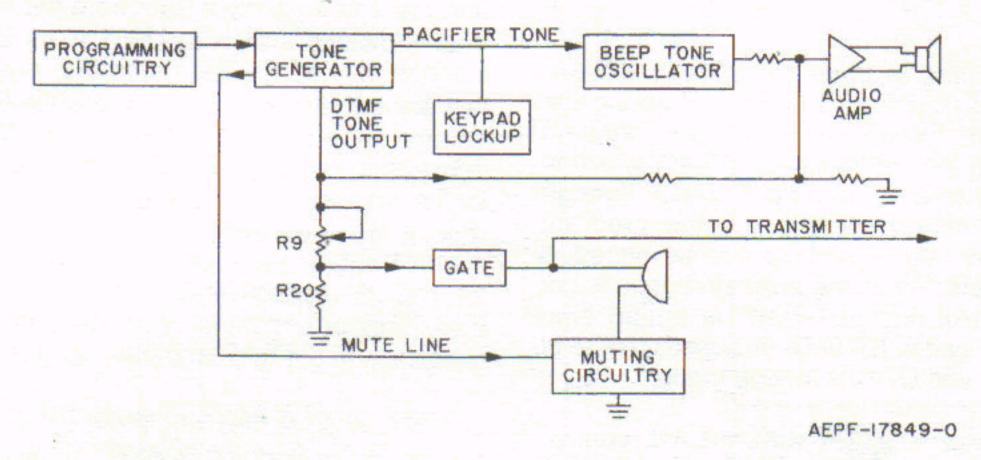


Figure 14. DTMF Option Diagram

circuit. Module U701, the heart of the DTMF circuit, receives its supply voltage (5V) through steering diode CR702. When the radio is turned off, the 5 volts is removed and U701 is supplied with memory retention voltage from the radio's unswitched B+ through resistor R701 and steering diode CR701. Resistor R706 and capacitor C701 act as a low-pass filter to keep noise off the IC's supply line. Capacitor C701 is also a memory retention cap. When the radio's battery is removed C701 will hold memory retention voltage for 2 minutes. If the battery is not replaced within two minutes, memory will be lost. All of the 47 pf caps are used for rf bypassing.

Transistor Q701, and resistors R722, R716 lock and unlock the keypad. When the radio is on, Q701 is saturated, U701 pin 22 is pulled low, and the keypad is unlocked. When the radio is turned off, Q701 is off, U701 pin 22 is pulled high through resistor R716, and the keypad is locked-up.

Integrated circuit U701 is a CMOS tone generator. Components Y701, R726, C722, and C723 form the oscillator circuit for the tone generator. When a key is pressed, U701 goes into the encode mode and outputs the appropriate tone on pin 21. Module U701 also sends a low (MUTE output) from pin 23 to NOR gate U703D pin 3. The tone (DTMF OUT) is routed through the deviation adjusting network of R720 and potentiometer R709, and applied to pin 6 of isolation switch U704B. If the control "C" input at U704B pin 4 is high, the switch closure is made and the DTMF tone output at U704B at pin 7 is applied to the radio's INT MIC IN line via connector plug P701 pin 3.

The purpose of the isolation gate (NOR gate U703D) is to prevent the transmission of beep tones. Therefore, the switch (U704B) will only close when a DTMF tone is to be transmitted, which is determined by a high output of NOR gate U703D at pin 4. This high output is achieved when both inputs are low. One input (pin 3) goes low everytime a DTMF tone is

generated. The other input (pin 2) goes low whenever the radio is in transmit, via the saturation of transistor Q703.

The function of FET transistor Q702 is to mute the microphone during tone transmission. If the microphone was not muted, noise could get mixed with the DTMF tones and prevent successful decoding. Transistor Q702 is controlled by the MUTE output (U701 pin 23). When no tone is present, the mute line is pulled high by resistor R728, transistor Q702 is on, and the microphone has a low impedance path to ground. When a DTMF tone is generated the mute line goes low, Q702 is turned off, and the microphone is no longer grounded. Therefore, the microphone is muted. It is also necessary to mute the microphone when beep tones are generated. When a beep tone is present, switch U704C closes and transistor Q702 turns off. The time that Q702 stays off is controlled by the RC network of C724 and R719.

The combination of U703B, U703C, U704A, R707, R708, C705,C703, and R723 is the beep-tone oscillator circuit. When a "\*" or a "#" command key is pressed, or when any key is pressed during the program mode, module U701 generates a pacifier tone. This tone, which lasts for approximately 30 milliseconds, is applied to beep tone gate U704A, which responds with a low output at pin 9. The low at U704A pin 9 is applied to the beep-tone oscillator (U703B pin 10), which responds by generating a 2000 Hz beep tone. The beep tone continues until U703B pin 10 goes high, which is determined by the RC network of C703 and R723 (approximately 57 milliseconds).

DTMF and beep tones are routed to the sidetone / beep-tone amplifer U702. This IC amplifies the tones and sends them to the speaker. Amplifier U702 is enabled when pin 1 is pulled low through CR704A or CR704B, which occurs when module U701 is in the program mode or when the radio is in transmit.

Resistor R702 and capacitor C702 control the duration of the DTMF tones and the rate the tones are generated during automatic dialing. Tone duration is set at 150 ms.

Program switching is done by U704D, U703A, U705, R703, R704, R705, and CR703. Pressing the program button puts the DTMF circuit in the program mode by grounding the contol line of U704D, which in turn causes pin 39 of U701 to be pulled high through resistor R705. With module U701 in the program mode (U701 pin 39 high), numbers can be stored in the memory registers. When the program button is not pressed, the control line of U704D is pulled high through R727, the switch (U704D) is closed, pin 39 of U701 is grounded, and U701 is in tone mode.

R731 is removed in radios with the ANI version DTMF circuit to prevent programming DTMF functions. In order to put module U701 into the program mode (a high at U701 pin 39), both inputs (pins 6 and 8) of U703A must be low. This can be accomplished only by using the ANI programming fixture to push the program button.

# (2) Continuous Tone Option

Integrated circuit U801 is a DTMF tone generator, which accepts inputs from the keypad. The option is supplied from the radio's 8-volt line. During tone generation, the IC outputs a high on its MUTE line (pin 8). This output mutes the microphone by saturating Q804 which turns off Q802, resulting in a high impedance path to ground for the microphone. The MUTE line also turns on Q801, which supplies a path to ground for the resistor divider network of R804 and R805.

The tone generator outputs a tone on pin 16 of U801. This tone level is reduced by R804 and R805, and is applied to the radio's MIC line to be transmitted. The tones are also divided by R806 and R807, and routed to the side-tone amplifier (U802). The amplified tones are then sent to the radio's speaker for user feedback.

The amplifier is enabled by the radio's MIC line. In the transmit mode, the MIC line is at 5 volts. This turns on Q803 and pulls pin 1 of U802 low, enabling the amplifier.

# I. Multicall 100,000 (100k) Circuitry (Optional)

The multicall circuit receives its power from an 8-volt regulator (U407) on the controller flex, via connector plug P701. When the radio is turned on, the regulated 8 volts is received by the multicall board and applied to a 5-volt regulator (U903) at pin 2. The regulated 5-volt output from U903 is applied to the microcomputer (U901 pin 21), to the watchdog timer (U902, pin 16), to the keypad, and throughout the multicall circuitry.

The multicall circuit is operated via an enable/disable switch (S901) and a multicall keypad. The keypad contains digits 0 through 9, the \* key, and the # key. All input lines from the keypad and the enable/disable switch to U901 (pins 23 through 31) are normally at 5 volts. Digits 0 through 9 and the \* key are connected to the microcomputer via row and column inputs. Whenever one of these keys is depressed, the corresponding row and column inputs to the microcomputer are grounded. The # key and the enable/disable switch are tied to the microcomputer via seperate sense lines (U901 pins 24 and 23 respectively). And likewise, when one of these (key or switch) is depressed, the corresponding sense line to the microcomputer is grounded.

The multicall microcomputer (U901)handles:

- all interface on the keypad
- radio control communications to the radio's microcomputer (U401) on the controller flex
- multicall system control

Module U901 reads in the keypad buttons depressed, and depending on the radio's codeplug programming, configures the radio's operation accordingly. Interface between the multicall's microcomputer (U901) and the radio's microcomputer (U401) is accomplished via control codes sent and received on a communications bus, which consists of DATA (U901 pin 10) and BUSY (U901 pin 7). Information is sent and received on the DATA line, while the BUSY line monitors the bus for use.

Timing for the multicall microcomputer is controlled by clock crystal Y901 and associated circuitry. The clock frequency is automatically shifted as a direct result of the receiver frequency. The voltage level on pin 33 of U901 controls the shifting network. If no frequency shift is required (as sensed by the microcomputer), 5 volts is applied to the base of transistor Q902, which turns on the transistor and shorts across coil L903. If a frequency shift is required, the microcomputer removes the 5 volts on pin 33, Q902 turns off, and L903 is placed in series with the crystal (Y901). The added inductance shifts the resonant frequency of the oscillator circuit, resulting a lower clock frequency.

To monitor the microcomputer's (U901) operation and to reset the microcomputer should the device fail, "get lost", or fall asleep during it's operation, a watchdog timer (U902) is incorporated in the multicall circuitry. The watchdog timer generates a 12.5 kHz pulse rate interrupt (wake up) signal (U902 pin 1 to U901 pin 18) to wake up the microcomputer if it is in the sleep mode. The microcomputer sends a return signal to reset the watchdog timer (U901 pin 9 to U902 pin 12). If the microcomputer should fail to send a reset signal to the watchdog timer, the watchdog timer sends a second interrupt signal to U901 pin 18.

If the microcomputer again fails to send a reset signal to the watchdog timer, the watchdog timer sends a pulse to Q901. Transistor Q901 inverts the pulse

and and applies it to U901 pin 22, which resets the microcomputer.

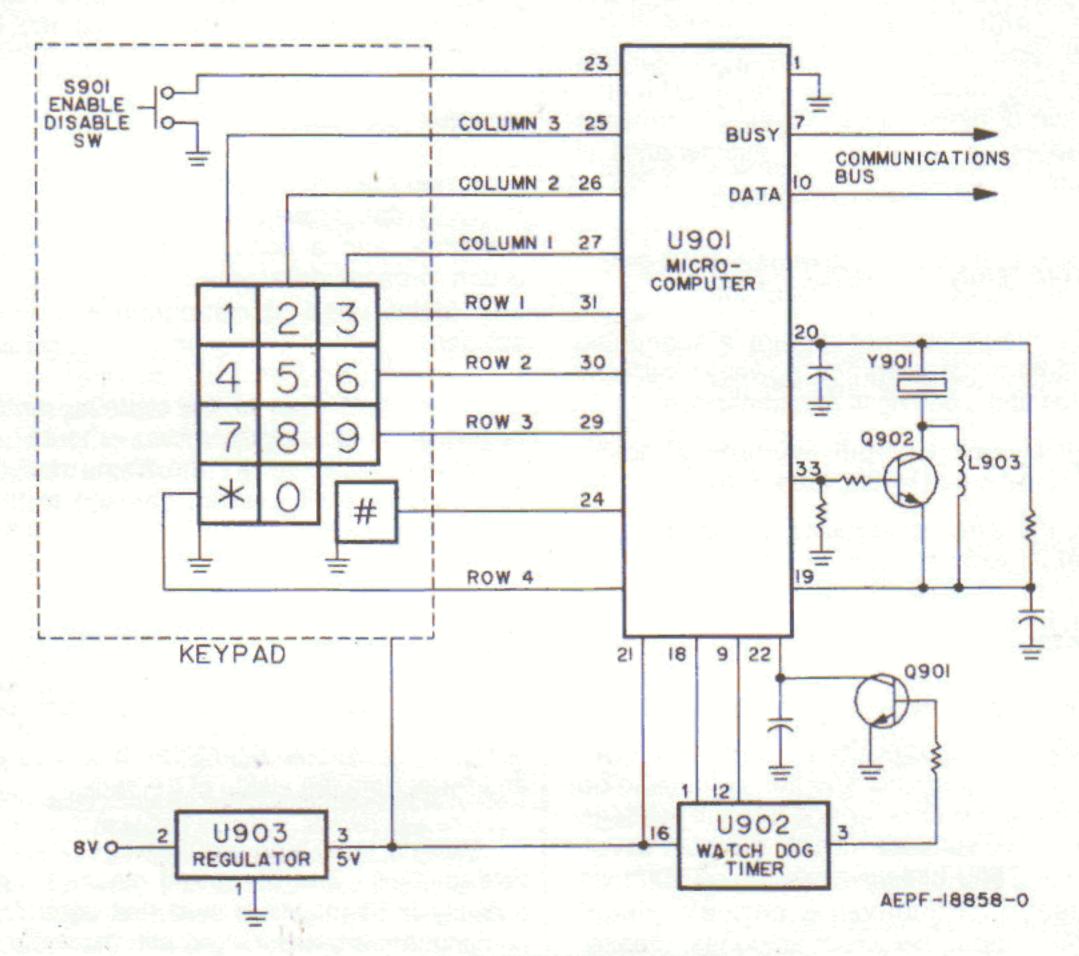


Figure 15. Simplified Multicall Circuitry

# MAINTENANCE

#### 1. INTRODUCTION

This section of the manual describes the disassembly and reassembly procedures, recommended repair procedures, special precautions regarding maintenance, and recommended test equipment. Each of these topics provides information vital to the successful operation and maintenance of the HT800 radio.

#### 2. PREVENTIVE MAINTENANCE

The HT800 radio does not require a scheduled preventive maintenance program; however, periodic visual inspection and cleaning is recommended.

#### a. Inspection

Check that the external surfaces of the radio are clean, and that all external controls and switches are functional. A detailed inspection of the interior electronic circuitry is not needed or desired.

#### b. Cleaning

The following procedures describe the recommended cleaning agents and the methods to be used when cleaning the external and internal surfaces of the radio. External surfaces include the front cover, housing assembly, and battery case. These surfaces should be cleaned whenever a periodic visual inspection reveals the presence of smudges, grease, and/or grime. Internal surfaces should be cleaned only when the radio is disassembled for servicing or repair.

The only recommended agent for cleaning the external radio surfaces is a 0.5% solution of a mild dishwashing detergent, such as JOY<sup>®</sup>, in water. The only factory recommended liquid for cleaning the printed circuit boards and their components is ISOPROPYL alcohol (70% by volume).

#### CAUTION

The effects of certain chemicals and their vapors can have harmful results on certain plastics. Aerosal sprays, tuner cleaners and other chemicals should be avoided.

#### (1) Cleaning External Surfaces

#### (a) Polycarbonate Surfaces

The detergent-water solution should be applied sparingly with a stiff, non-metallic, short-bristled brush

to work all loose dirt away from the radio. A soft, absorbent, lintless cloth or tissue should be used to remove the solution and dry the radio. Make sure that no water remains entrapped near the connectors, cracks, or crevices.

#### (b) Silverized Surfaces

A non-metallic, soft-bristled brush should be used to apply the detergent-water solution to silverized surfaces, and a second non-metallic soft-bristled brush (free of detergent or rinsed in clean water) should be used to remove the detergent-water solution.

Upon completion of the cleaning process, a soft, absorbent, lintless cloth or tissue should be used (with a blotting action) to dry the frame and covers. The blotting action will prevent damage to the silverized conductive coating.

### (2) Cleaning Internal Circuit Boards and Components

Isopropyl alcohol may be applied with a stiff, nonmetallic, short-bristled brush to dislodge embedded or caked materials located in hard-to-reach areas. The brush stroke should direct the dislodged material out and away from the inside of the radio.

Alcohol is a high-wetting liquid and can carry contamination into unwanted places if an excessive quantity is used. Make sure that controls or tunable components are not soaked with the liquid. Do not use high-pressure air to hasten the drying process, since this could cause the liquid to puddle and collect in unwanted places.

Upon completion of the cleaning process, use a soft, absorbent, lintless cloth to dry the area. Do not brush or apply any isopropyl alcohol to the frame, front cover, or back cover.

#### NOTE

Always use a fresh supply of alcohol and a clean container to prevent contamination by dissolved material (from previous usage).

# 3. DISASSEMBLY (Refer to the exploded view in the HT800 service manual)

Disassembly of the radio involves removal of the major components listed below, one at a time, in the sequence described in the following paragraphs.

#### NOTE

- Several special tools are required to completely disassemble the radio. Refer to the "Test Equipment and Service Aids" paragraphs in this section of the manual. Also refer to the "Torque and Tool Specifications Chart."
- Before proceeding, make sure that the radio is turned off.

## a. Battery Removal

To remove the battery from the radio, proceed as follows:

- Step 1. Hold the radio with the front of the radio facing up.
- Step 2. Disengage the battery latch from the battery by pushing and holding the latch towards the top of the radio.
- Step 3. With the battery latch disengaged, slide the battery from left to right to remove it from the baseplate on the bottom of the radio housing.

## b. Gaining Access to Internal Components

#### CAUTION

The HT800 radio contains complementary metal-oxide semiconductor (CMOS) devices, which are highly susceptible to damage in handling due to static discharge. The entire printed circuit board should be treated as static sensitive. Damage can be latent, resulting in failures occurring weeks or months later.

DO NOT attempt to disassemble the radio without first referring to the "Safe Handling of CMOS Devices" paragraph in this section of the manual.

- Step 1. Remove the battery as described in paragraph a.
- Step 2. Remove the two screws from the back of the radio.
- Step 3. Remove the two screws on the bottom of the radio (baseplate corners).
- Step 4. Lift the front cover from the radio housing, being careful not to pull against the speaker/microphone flex.
- Step 5. Disconnect the speaker/microphone connector from the controller flex by grasping the speaker flex strain relief (near the plug) and pulling the plug straight out and away from the circuit board.

- Step 6. Loosen the two captive screws on the bottom of the radio. Do not completely remove the captive screws from the baseplate.
- Step 7. With a thumb and forefinger, grasp the antenna at its base and pull lightly to remove the frame assembly from the radio housing. Do not press the PTT switch during removal.
- Step 8. Remove the antenna by unscrewing it counterclockwise.
- Step 9. Remove the screw that secures the front shield to the controller carrier.
- Step 10. Remove the front shield by pulling it straight out and away from the radio.
- Step 11. Remove the four screws that secure the main back shield to the frame.
- Step 12. Remove the main back shield by pulling it straight out and away from the radio.

# c. Removing the Controller Assembly

- Step 1. Perform steps 1 through 10 of paragraph b.
- Step 2. Remove the four screws (two on each side) that secure the controller carrier to the frame.

#### NOTE

Be careful to pull each connector straight out and away from the mating socket so as not to bend or break the connector pins.

- Step 3. Disconnect the two bottom flex connectors by carefully sliding them away from the synthesizer.
- Step 4. Lift the controller circuit (nearest the bottom of the radio) away from the radio just enough to gain access to the connector under the controller.
- Step 5. Disconnect the connector under the controller.
- Step 6. Disconnect the two connectors at the top of the controller.
- Step 7. Lift the controller assembly totally away from the radio.

#### d. Gaining Access to the Controller Flexible Circuit

- Step 1. Perform steps 1 through 7 of paragraph c.
- Step 2. Remove the screws that secure the bottom shield to the top flex carrier.
- Step 3. Along the top edge of the controller assembly (edge nearest speaker clearance indentation), gently pry the bottom shield away from the top flex carrier.
- Step 4. Pull the bottom shield completely away from the top flex carrier and remove the controller flexible circuit.

# e. Removing the Transceiver Board from the Frame

- Step 1. Perform steps 1 through 7 of paragraph c.
- Step 2. Remove the four screws that secure the main back shield, and remove the shield.
- Step 3. Unsolder four contacts (two pins and one frame ground connection) located next to the screw (back, top-center of transceiver board), and the antenna ferrule located on the back, top-left corner of the transceiver board.
- Step 4. Remove one screw (back, top-center of transceiver board) that secures the transceiver board to the frame.
- Step 5. Unsolder and remove the red B+ wire (controller side of radio) from the On-Off / volume switch pot.

#### CAUTION

Always place the On-Off / Volume switch pot in the 'On' position before soldering to this switch, and return to the 'Off' position when finished soldering.

Step 6. Gently pull the transceiver circuit board straight out and away from the frame.

# f. Removing the Control-Top Panel Components

Step 1. Perform steps 1 through 5 of paragraph e.

#### NOTE

All control-top panel components, except the antenna jack, are connected on two flexible circuits, which are connected together and should be removed as one unit.

Step 2. Remove the control knob(s) by pulling straight out and away from the controltop panel.

- Step 3. Remove the teflon washer(s).
- Step 4. The escutcheon is stuck to the top surface of the control-top panel with adhesive. Gently pry one corner of the escutcheon away from the control-top panel and then peel the escutcheon completely away. Notice that washer(s) are stuck on the back side of the escutcheon.
- Step 5a. Remove the hex nut and washer from the volume potentiometer.
- Step 5b. Remove the hex nut and washer from the frequency switch.
- Step 6. Remove the spanner nut and washer from the PL switch.
- Step 7. Pry the header (part of volume pot flex assembly) away from the universal connector pins.
- Step 8. Unsolder the three legs of the LED and pull the flex away from the legs.
- Step 9. Unsolder and remove the black wire (ground wire from header to frame) where it contacts the frame.
- Step 10. The frequency switch flex connects to the PTT / B+ flex with five solder tabs located along the side of the frame near the monitor popple switches. Unsolder the five contact tabs, and with "solder- wick", remove the solder and separate the two flexes.

#### NOTE

A capacitor is placed across the last two tabs.

Step 11. Push the switch shaft(s) until clear of the mounting holes, and remove the flex circuits and control-top panel components away from the frame.

#### g. Removing the Control-Top Panel and LED

- Step 1. Perform steps 1 through 8 of paragraph f.
- Step 2. Unsolder the ground pin of the universal connector contacting the frame (near the antenna bushing).
- Step 2a. Remove the screw and washer located near the antenna receptacle.
- Step 2b. Gently pull the control-top panel away from the frame.
- Step 2c. Push the LED and rubber boot out of the control-top panel, and pull the LED out of the rubber boot.

## h. Removing the Battery Latch

- Step 1. Perform steps 1 through 7 of paragraph b.
- Step 2. Remove the ground contact screw that holds the negative battery contact. Be careful not to lose the lockwasher, contact, and rubber pad (under the contact).
- Step 3. While holding the latch slide, carefully pull the baseplate assembly away from the housing.
- Step 4. Carefully slide the latch out of the housing.
- Step 5. Remove the exposed latch springs.

#### i. Removing the PTT / B+ Flex

- Step 1. Perform steps 1 through 7 of paragraph b.
- Step 2. Two corners of the PTT / B+ flex are soldered to the frame. Remove the solder, using "solder-wick".
- Step 3. The PTT / B+ flex connects to the frequency switch flex with five solder tabs located along the top side of the frame near the PL switch. Unsolder the five contact tabs, and with "solder-wick", remove the solder and separate the two flexes.

NOTE

A capacitor is placed across the last two tabs.

- Step 4. The PTT / B+ flex is stuck to the frame with adhesive. Carefully peel the PTT / B+ flex away from the frame.
- 4. REASSEMBLY (Refer to the exploded view in the HT800 service manual.)
  - DO NOT attempt to reassemble the radio without first referring to the "Safe Handling of CMOS Devices" paragraph in this section of the manual.
  - DO NOT attempt to reassemble the radio without first referring to the approriate VHF or UHF service manual "TORQUE AND TOOL SPECIFICATIONS CHART".
  - Inspect all O-rings and replace if obvious damage exists.

### a. Reinstalling the Battery Latch and Base Plate

Step 1. Insert the two springs into their proper holes, and replace the slide latch.

- Step 2. Position the base plate and hold it firmly to compress the springs.
- Step 3. Holding the base plate in place, install the negative battery contact, being sure that the rubber pad is in place in the cup of the contact.
- Step 4. Reinstall the screw and lockwasher in the negative battery contact. Tighten the screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".

## b. Reinstalling the PTT / B+ Flex

- Step 1. Position the PTT / B+ flex to the frame such that the five contact tabs line up with the corresponding tabs on the frequency switch flex. Note that a little oval hole in the corner of the flex (near the solder tabs) mates with a round dot on the frame.
- Step 2. Press the flex to the frame. Note that two more places, holes in the flex correspond with dots on the frame.
- Step 3. Resolder the five solder tabs connecting the PTT / B+ flex to the frequency switch flex.

NOTE

A capacitor is placed across the last two tabs.

Step 4. Resolder the two corners of the flex to the frame.

#### c. Reinstalling the LED and Control-Top Panel

- Step 1. Insert the LED into the rubber boot such that the flat edge of the LED's base mates with the flat edge inside the boot.
- Step 2. Insert the LED and boot into the controltop panel.
- Step 3. Place the control-top panel on the frame.
- Step 4. Reinstall the screw and washer located near the antenna receptacle, and tighten the screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 5. Resolder the ground pin of the universal connector to the frame.

#### d. Reinstalling the Control-Top Panel Components

- Step 1. Insert the switch shafts into the proper holes.
- Step 2. Resolder the three LED legs to the frequency switch flex.

- Step 3. Press the volume pot header on to the corresponding pins of the universal connector.
- Step 4. Resolder the black ground wire to the frame.
- Step 5. Resolder the five solder tabs of the frequency switch flex to the corresponding tabs of the PTT / B+ flex.
- Step 6. Reinstall the PL switch washer and spanner nut, and tighten per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 7. Reinstall the frequency switch and volume pot washers and hex nuts, and tighten each screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 8. Reinstall the escutcheon.
- Step 9. Reinstall the teflon washers on the frequency switch and volume pot shafts.
- Step 10. Reinstall the switch knobs.

# e. Reinstalling the Transceiver Board

- Step 1. With the frame's backside laying down, and viewing the transceiver board from the solder side with the assembly upright, slightly spread the sides of the frame and slide the transceiver into the frame.
- Step 2. Turn the unit over and resolder the loose end of the red B+ wire to the On-Off/Volume switch pot.

#### CAUTION

Always place the On-Off / Volume switch pot in the 'On' position before soldering to this switch, and return to the 'Off' position when finished soldering.

- Step 3. Reinstall one screw (back, top-center of transceiver board) that secures the transceiver board to the frame, and tighten securely.
- Step 4. Resolder four contacts (two pins and one frame ground connection) located next to the screw (back, top-center of transceiver board), and the antenna ferrule contact (back top-left corner of board).
- Step 5. Press the main back shield (edges over the frame) flush to the transceiver board.
- Step 6. Reinstall the four screws that secure the main back shield to the frame, and tighten

each screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".

# g. Reassembling the Controller Assembly

#### CAUTION

Make sure that the flex insulator is installed around the controller flex before placing the controller flex into the carrier.

- Step 1. With the outside surface of the carrier laying down, and the controller flex folded over (shield-to-shield), align the holes in the flex with corresponding holes in the carrier, and place the flex into the carrier. Make sure that the P1 and P2 jack's grooves slide into the tabs of the carrier. Also, make sure that the J5 jack is seated properly in the carrier.
- Step 2. Align the controller bottom shield to the controller flex and carrier. In the J5 jack area, slide the tab of the shield under the slot in the carrier, and press the bottom shield into place (sides of the bottom shield fit inside the sides of the carrier).
- Step 3. Reinstall the screws that secure the bottom shield to the controller carrier, and tighten each screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".

# g. Reinstalling the Controller Assembly

#### NOTE

Be careful to push each connector straight into the mating socket so as not to bend or break the connector pins.

- Step 1. Reconnect the two top flex connectors, firmly seating both plug / jack connections.
- Step 2. Reconnect the connector under the controller, firmly seating the plug / jack connection.
- Step 3. Press the controller into place (inside of frame sides).
- Step 4. Reconnect the two bottom flex connectors, firmly seating both plug / jack connections.
- Step 5. Reinstall the four screws (two on each side) that secure the controller carrier to the frame, and tighten each screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 6. Reinstall front shield (shield edges fit inside the frame).

Step 7. Reinstall the screw that secures the front shield to the controller carrier, and tighten the screw per the "TORQUE AND TOOL SPECIFICATIONS CHART".

# h. Final Reassembly

- Step 1. Insert the internal radio unit into its housing, and tighten the two screws on the baseplate per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 2. Reconnect the speaker / microphone connector, being careful to push the connector straight into the mating socket so as not to bend or break the connector pins.
- Step 3. Reinstall the front cover.
- Step 4. Reinstall the two screws on the bottom of the radio (baseplate corners), and tighten the screws per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 5. Reinstall the two screws that secure the front cover to the housing, and tighten each screws per the "TORQUE AND TOOL SPECIFICATIONS CHART".
- Step 6. Reinstall the antenna.
- Step 7. Reinstall the battery.

#### 5. SAFE HANDLING OF CMOS DEVICES

Complementary metal-oxide semiconductor (CMOS) devices are used in the HT800 radio. While the attributes of CMOS are many, their characteristics make them susceptible to damage by electrostatic or high voltage charges. Damage can be latent, resulting in failures occurring weeks or months later. Therefore, special precautions must be taken to prevent device damage during disassembly, troubleshooting, and repair. The following handling precautions are mandatory for CMOS circuits, and are especially important in low humidity conditions.

- a. All CMOS devices must be stored or transported in conductive material so that all exposed leads are shorted together. CMOS devices must not be inserted into conventional plastic "snow" or plastic trays of the type that are used for storage or transportation of other semiconductor devices.
- b. All CMOS devices must be placed on a grounded bench surface and the technicians must ground themselves prior to handling the devices. This is done most effectively by having the technician wear a conductive wrist strap in series with a 1M ohm resistor to ground.

- c. Do not wear nylon clothing while handling CMOS circuits.
- d. Do not insert or remove CMOS devices with power applied. Check all power supplies to be used for testing CMOS devices, and be certain that there are no voltage transients present.
- When straightening CMOS device leads, provide ground straps for the apparatus used.
- f. When standing, use a grounded soldering iron.
- g. All power must be turned off in a system before printed circuit boards containing CMOS devices are inserted, removed, or soldered.

#### 6. REPAIR PROCEDURES AND TECHNIQUES

#### CAUTION

Leadless component technology requires the use of specialized equipment and procedures for repair and servicing of the HT800 radio. If you are not totally familiar with leadless component repair techniques, it is strongly recommended that you either defer maintenance to qualified service personnel and service shops or take the recommended video taped leadless component repair training program, MAV-PACK 3 (VID-952) (see paragraph 7b, Service Aids and Recommended Tools, in this section). This is of paramount importance as irreparable damage to the radio can result from service by unauthorized persons. Unauthorized attempts to remove or repair parts may void any existing warranties or extended performance agreements with the manufacturer.

# a. Parts Replacement and Substitution

Special care should be taken to be as certain as possible that a suspected component is actually the one at fault. This special care will eliminate unnecessary unsoldering and removal of parts, which could damage or weaken other components or the printed circuit board itself.

When damaged parts are replaced, identical parts should be used. If the identical replacement component is not locally available, check the parts list for the proper Motorola part number and order the component from the nearest Motorola Communications Parts office listed in the "Replacement Parts Ordering" section of this manual.

#### b. Rigid Circuit Boards

The HT800 radio uses bonded multi-layer printed circuit boards. Since the inner layers are not accessible, some special considerations are required when soldering and unsoldering components. The printed through holes may interconnect multiple layers

of the printed circuit. Therefore, care should be exercised to avoid pulling the plated circuit out of the hole.

When soldering near the module socket pins, use care to avoid accidentally getting solder in the socket. Also, be careful not to form solder bridges between the module socket pins. Closely examine your work for shorts due to solder bridges. When removing modules with metal enclosures, be sure to desolder the enclosure ground tabs as well as the module pins.

#### c. Flexible Circuits

The flexible circuits are made from a different material than the rigid boards, and different techniques must be used when soldering. Excessive prolonged heat on the flexible circuit can damage the material. Avoid excessive heat and excessive bending. For parts replacement, use the ST-1087 Temperature-Controlled Solder Station with a 600 or 700 degree tip, and use small diameter solder such as ST-633. The smaller size solder will melt faster and require less heat being applied to the circuit.

To replace a component on a flexible circuit, grasp the edge of the flexible circuit with seizers near the part to be removed, and pull gently. Apply the tip of the soldering iron to the component connections while pulling with the seizers. Do not attempt to puddle out components. Prolonged application of heat may damage the flexible circuit.

#### 7. TEST EQUIPMENT AND SERVICE AIDS

The following paragraphs describe the test equipment and service aids required for maintaining the HT800 radio. Your Motorola sales representative will assist in analyzing your specific requirements and help you select the latest available equipment to suit your individual needs. In addition, your sales representative can advise you of the availability of new test equipment and service aids that become available after the printing of this manual.

Refer to Figure 16 for an illustration of troubleshooting, test equipment, and programming set-up.

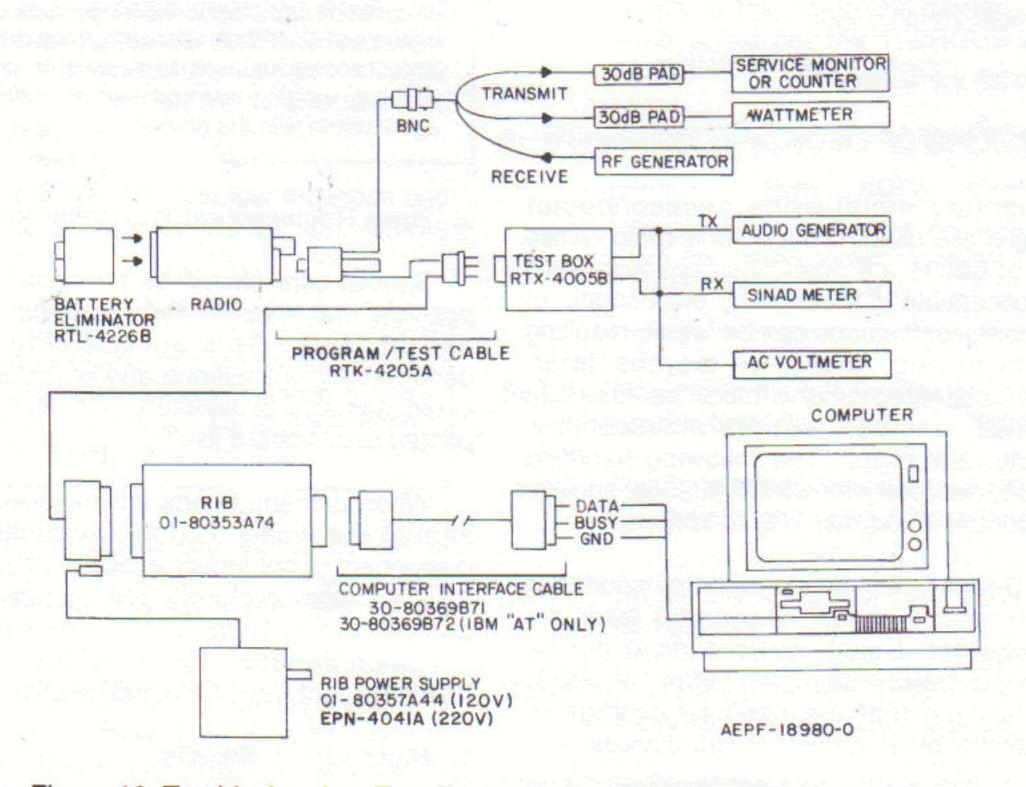


Figure 16. Troubleshooting, Test Equipment, and Programming Set-Up Detail

#### a. Recommended Test Equipment

The list of equipment contained in Table 2 includes all of the standard test equipment required for servicing two-way portable radios, as well as several unique items designed specifically for servicing the HT800 radio. Battery operated test equipment is recommended when available. The "Characteristics" column is included so that equivalent equipment may be substituted; however, when no information is provided in this column, the specific Motorola model listed is either a unique item or no substitution is recommended.

Table 2. Test Equipment

MOTOROLA MODEL NUMBER	DESCRIPTION	CHARACTERISTICS	APPLICATION	
R2200, R2400, or R2001D with trunking option	Service Monitor	This monitor will substitute for items with an asterisk (*)	Frequency/deviation meter and signal generator for wide-range troubleshooting and alignment	
*R1049A	Digital Multimeter		Two meters recommended for ac/dc voltage and current measurements	
*S1100A	Audio Oscillator	67 to 161.4Hz tones	Used with service monitor for injection of PL tones	
*S1053D *SKN6009A *SKN6001A	AC Voltmeter Power Cable for Meter Test Leads for Meter	1mV to 300V, 10-Megohm input impedance	Audio voltage measurements	
R1053	Dual-Trace Oscilloscope	20MHz bandwidth 5mV/cm - 20V/cm	Waveform measurements	
*S1350C *ST1215B (VHF) *ST1223B (UHF) *T1013A	Watt Meter Plug-in Element RF Dummy Load	50-ohm, ±5% accuracy 10 Watts, maximum 0-1000MHz, 300W	Transmitter power output measurements	
S1339A	RF Millivolt Meter	100μV to 3V rf 10kHz to 1.2GHz	RF level measurements	
*R1013A	SINAD Meter		Receiver sensitivity measurements	
S1347D or S1348D (programmable)	DC Power Supply	0-20Vdc, 0-5 Amps* current limited	Bench supply for 7.5Vdc	
R1150B	SELECT 5 Encoder/Decoder	Encodes and decodes 681-2800Hz tones	SELECT 5 Signalling testing and measurements	

R2200, R2400, or R2001D will substitute for items with an asterisk (\*)

#### b. Service Aids and Recommended Tools

Refer to the appropriate VHF or UHF service manual ("SERVICE AIDS" and "RECOMMENDED TOOL LIST") for a listing and description of the service aids and tools designed specifically for servicing the HT800 radio, as well as the more common tools required to disassemble and properly maintain the radio. These kits and/or parts are available from the Motorola Communications Parts office listed in the "Replacement Parts Ordering" section of this manual.

#### MAV-PACK 3 (VID-952)

The VID-952 Motorola Video Visual Package (MAV-PACK) is a video tape training program on leadless component repair techniques. This VHS format video cassette and supplemental literature describe the removal and replacement of leadless components using the following specialized equipment:

- RRX-4033 Laurier Hot Gas Bonder
- RPX-4234A Regulator and Hardware Kit

- 0180386A62 Heated Tweezers
- RSX-1002 Desoldering Station
- RSX-1008 Weller Soldering Station

This MAV-PACK is strongly recommended for technicians who intend to service this and other Motorola radios using leadless components. This VHS videotape is in standard half-inch format. This MAV-PACK, as well as others, is available from:

Motorola C&E, Inc. National Service Training Center 1300 N. Plum Grove Road Schaumburg, Illinois 60195

#### 8. FIELD PROGRAMMING

The HT800 radio can be field programmed. Field programming requires specific equipment and accompanying instructions. Refer to the HT800 "Radio Service Software User's Manual" (Motorola number 68P02000F01) for complete field programming information.

# TROUBLESHOOTING

#### 1. INTRODUCTION

Servicing the HT800 Series radio requires the localization of the malfunctioning circuit before the defective component can be isolated and replaced. Since localizing and isolating a defective component constitutes the most time consuming part of troubleshooting, a thorough understanding of the circuits involved will aid the technician in performing efficient servicing. The technician must know how one function affects another; he must be familiar with the overall operation of the radio and the procedures necessary to place it back in operation in the shortest possible time.

The radio functional block diagrams, schematic diagrams, and troubleshooting charts provide valuable information for troubleshooting purposes. The functional diagrams provide signal flow information in a simplified format, while the schematic diagrams provide the detailed circuitry and the biasing voltages required for isolating malfunctioning components. By using the diagrams, troubleshooting charts, and deductive processes, the suspected circuit may be readily found.

To determine if analyzation of the radio is required, perform checks such as 20dB SINAD psophometrically weighted, 12dB SINAD, noise and PL squelch sensitivity, for the receiver; and current drain for the transmitter. These should give the technician a general indication of where the problem is located.

### NOTE

See Figure 16 - Troubleshooting, Test Equipment, and Programming Set-Up Detail.

After the general problem area of the radio has been identified, careful use of a dc voltmeter, rf millivoltmeter, and an oscilloscope should isolate the problem to an individual component.

#### 2. TROUBLESHOOTING PROCEDURE

Each time that the radio is turned on, a microcomputer self-test occurs. A 1600Hz alert tone is generated for approximately 500 milliseconds to indicate that the microcomputer is functioning properly. If the alert tone is not heard (and the alert tones have not been disabled via the Radio Service Software), there is a problem with the radio.

Following the microcomputer self-test, a synthesizer self-test occurs. A continuous 1600Hz alert tone is generated if the synthesizer test is **not** successful. If this condition occurs (continuous alert tone) refer to the VCO/synthesizer troubleshooting chart.

When a radio performs unsatisfactorily, the following procedures should help localize the fault.

#### a. Check Batteries

The first step in localizing a trouble is to check the battery voltage under load. With the transmitter turned on (keyed), check the battery voltage. A convenient way to do this is to remove the front cover and monitor the B+ line with a voltmeter (with respect to ground). The measured load voltage should not be less than eight volts. Even though the transmitter may operate at a lower voltage, operation would be marginal and for only a short period of time. Low-voltage transmit operation is indicated by the flashing LED on top of the radio. If the measured voltage is zero volts, check the battery and fuse. The recommended procedure is to replace, or recharge, the battery if the voltage is below eight volts under load.

# b. Alignment

Strict adherence to the published procedures is a prerequisite to accurate alignment and proper evaluation of the performance of the radio. The selection of test equipment is critical. The use of equipment other than that recommended should be cleared through your Motorola Area Representative to ensure that it is of equivalent quality.

The service technician must observe good servicing techniques. The use of interconnecting cables that are too long, poorly positioned (dressed), or improperly terminated will result in erratic meter readings. As a result, it will not be possible to tune the radio to the desired specifications.

Use the recommended test equipment setup and proper connections for alignment and adjustments. Refer to the detailed procedures supplied in the applicable service manual.

#### c. Check Overall Transmitter Operation

If the battery voltage is sufficient, check the overall performance of the transmitter. A good overall check of the transmitter is the rf power output measurement. This check indicates the proper operation of the transmitter amplifier stages. A properly tuned and operating transmitter will produce the rated rf output into a 50-ohm load with a dc input of 10 volts (refer to "Transmitter Alignment Procedure," located in the service manual, for specific rf output). If the power is less than rated rf output, refer to the applicable transmitter troubleshooting chart.

# d. Check Overall Receiver Operation

# (1) 20dB SINAD Psophometrically Weighted

This procedure is a standard method for evaluating the performance of an FM receiver, since it provides a check of the rf, i-f, and audio stages. The method consists of finding the lowest modulated signal necessary to produce 50% of the radio's rated audio output with a 20dB or better ratio of signal + noise + distortion / noise + distortion. This is termed "usable sensitivity."

To perform this measurement, connect the leads from a SINAD meter, which has a psophometric filtering network (CCIT filter), to the audio output of the test box. Set the Motorola service monitor or rf signal generator to output a 1-millivolt signal. Modulate the rf signal with a 1kHz tone at 60% of the total system deviation. Introduce the signal to the radio at the exact channel frequency through the universal connector. Set the volume control for half the rated audio output (3.16Vrms). Decrease the rf signal level until the SINAD meter reads 20dB. The signal generator output (20dB SINAD measurement) should be less than 0.50µV on both uhf and vhf receivers. If the radio does not meet this specification, try to retune the receiver using the procedure indicated in the service manual. If this does not solve the problem, refer to the receiver troubleshooting chart.

# (2) 12dB SINAD

This procedure is a standard method for evaluating the performance of an FM receiver, since it provides a check of the rf, i-f, and audio stages. The method consists of finding the lowest modulated signal necessary to produce 50% of the radio's rated audio output with a 12dB or better ratio of signal + noise + distortion / noise + distortion. This is termed "usable sensitivity."

To perform this measurement, connect the leads from a SINAD meter to the audio output of the test box. Set the Motorola service monitor or rf signal generator to output a 1-millivolt signal. Modulate the rf signal with a 1kHz tone at 60% of the total system deviation. Introduce the signal to the radio at the exact channel frequency through the universal connector. Set the volume control for half rated audio output (3.16Vrms). Decrease the rf signal level until the SINAD meter reads 12dB. The signal generator output (12dB SINAD measurement) should be less than 0.40μV on VHF receivers or less than 0.42μV on UHF receivers. If the radio does not meet this specification, try to retune the receiver using the procedure indicated in the service manual. If this does not solve the problem, refer to the receiver troubleshooting chart.

# 3. VOLTAGE MEASUREMENT AND SIGNAL TRACING

To aid in troubleshooting, ac and dc voltage readings are provided (in red) on the transceiver schematic diagram in the service manual. When making these voltage checks, pay particular attention to any notes that may accompany the voltage reading of a particular stage.

If receiver sensitivity is high or if the rf power output is lower than normal for a fully tuned transceiver, the dc voltages on the printed circuit board should be checked. These voltages should be referenced to ground.

### CAUTION

When checking a transistor or module, either in or out of the circuit, do not use an ohmmeter having more than 1.5 volts dc appearing across the test leads or an ohms scale of less than x 100.

It is recommended not to replace a transistor or module before a thorough check is made. Read the voltages around the suspected stage. If these voltages are not reasonably close to those specified, the associated components should be checked.

A low impedance meter should not be used for measurement. If all dc voltages are correct, the signal should be traced through the circuit to show any possibility of breaks in the signal path.

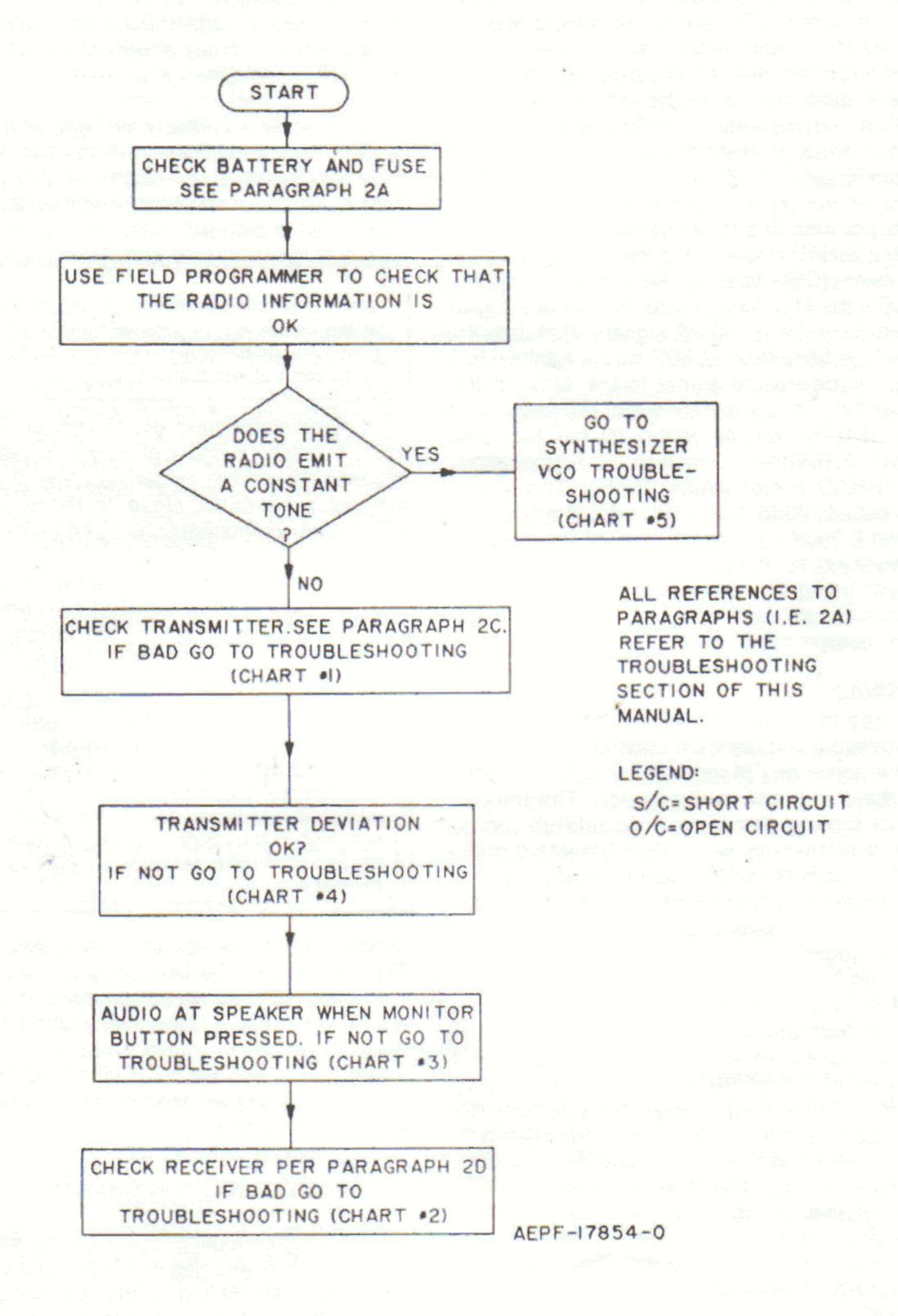
# CAUTION

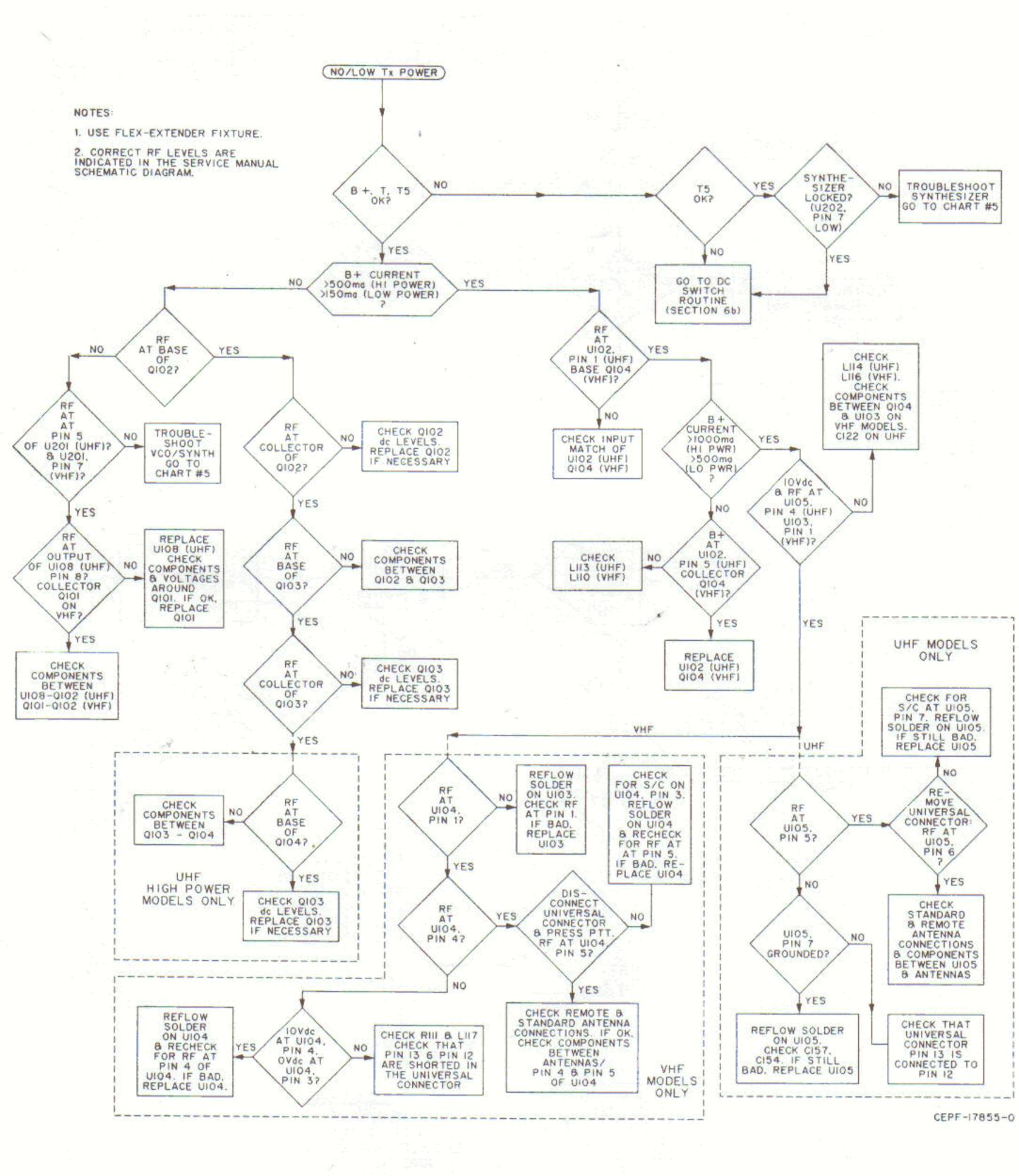
The microcomputer is a static sensitive device contained on the controller flex assembly. DO NOT attempt to troubleshoot or disassemble the microcomputer/controller flex assembly without first referring to the "Safe Handling of CMOS Devices" paragraph in the MAINTENANCE section of the manual.

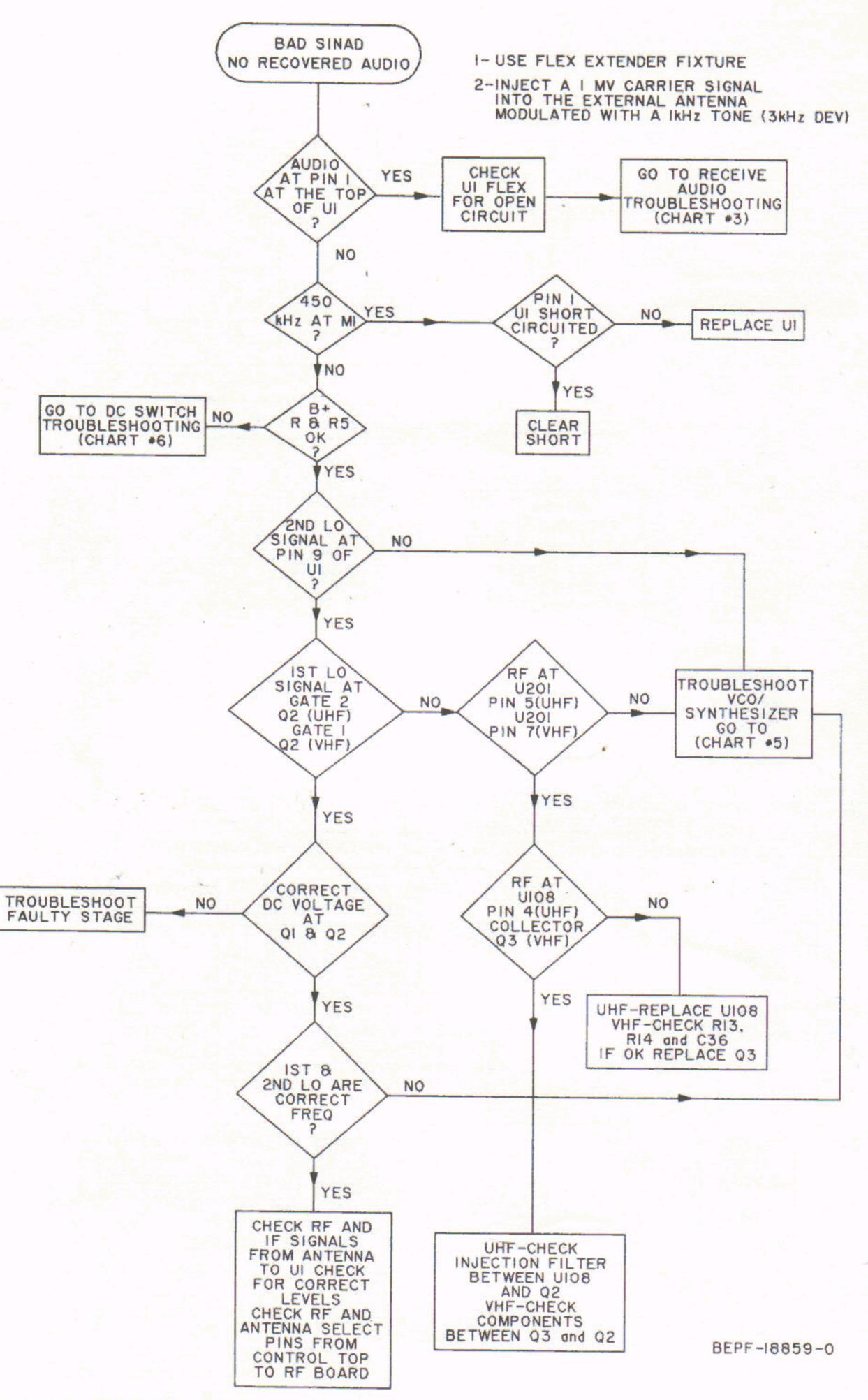
When troubleshooting the microcomputer controller flex circuits, it will be necessary to disconnect the flex from the radio main circuit board and reconnect it via a flex extender fixture. Also, many of the measurements referred to in the microcomputer troubleshooting charts that follow are short in duration. So, it will be necessary to use an oscilloscope set for 1V / division and 5ms / division.

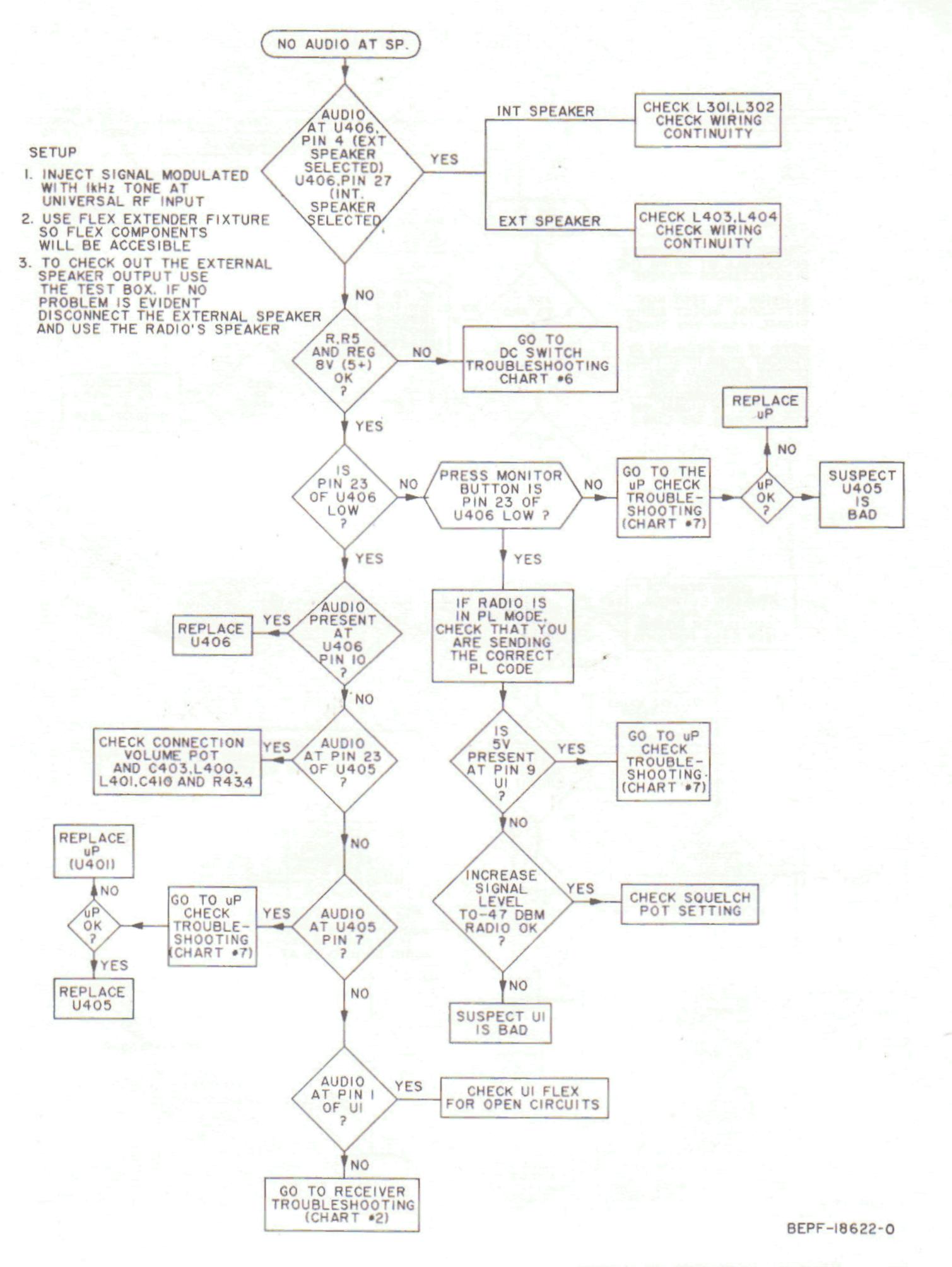
# 4. TROUBLESHOOTING CHARTS

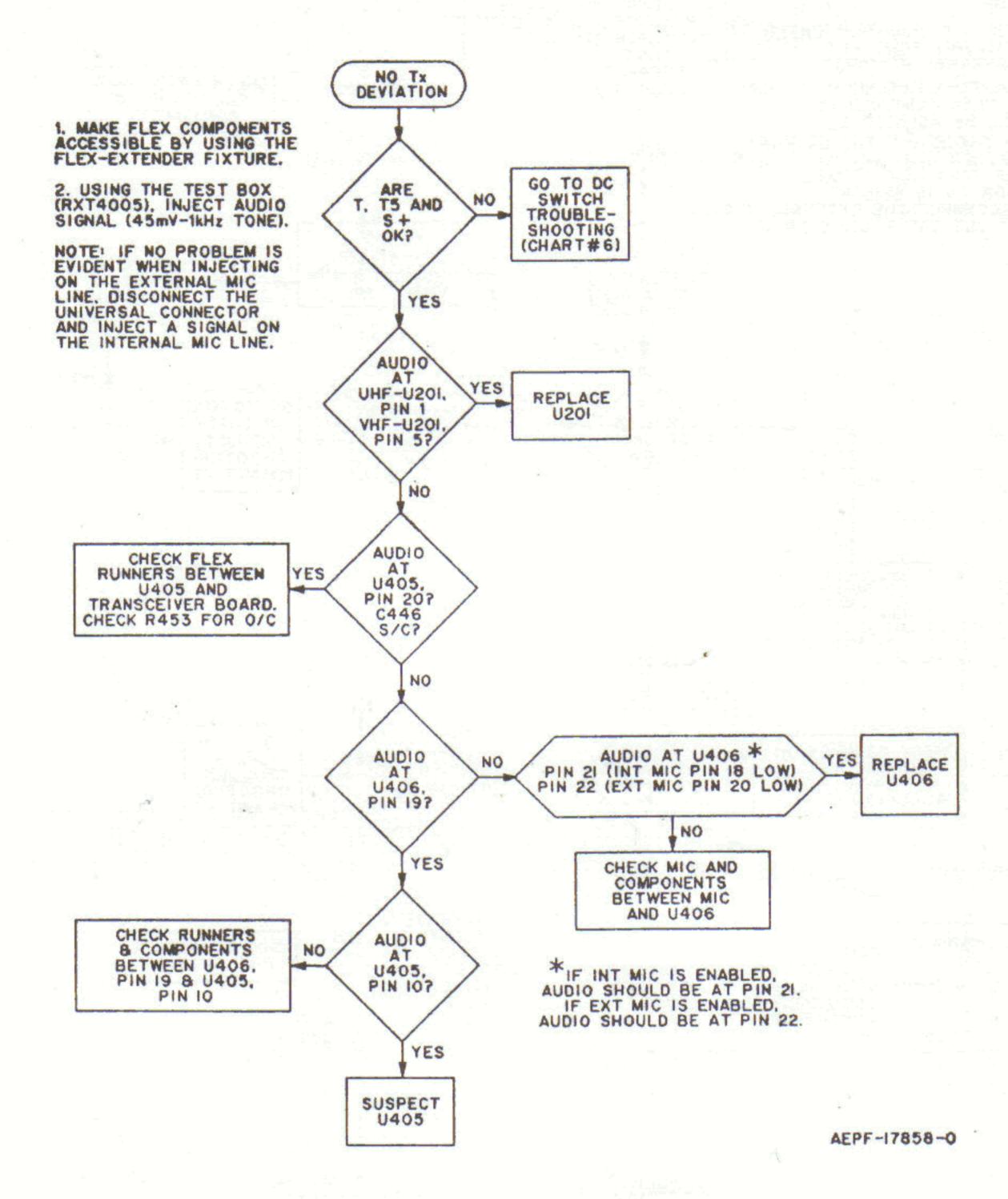
The troubleshooting charts on the following pages will help isolate troubles in the different sections of the radio. Start at the top of the appropriate chart and make the checks as indicated. Most usual malfunctions will respond to the systematic approach to troubleshooting. Also, a flowchart is provided to aid in choosing the proper troubleshooting chart.

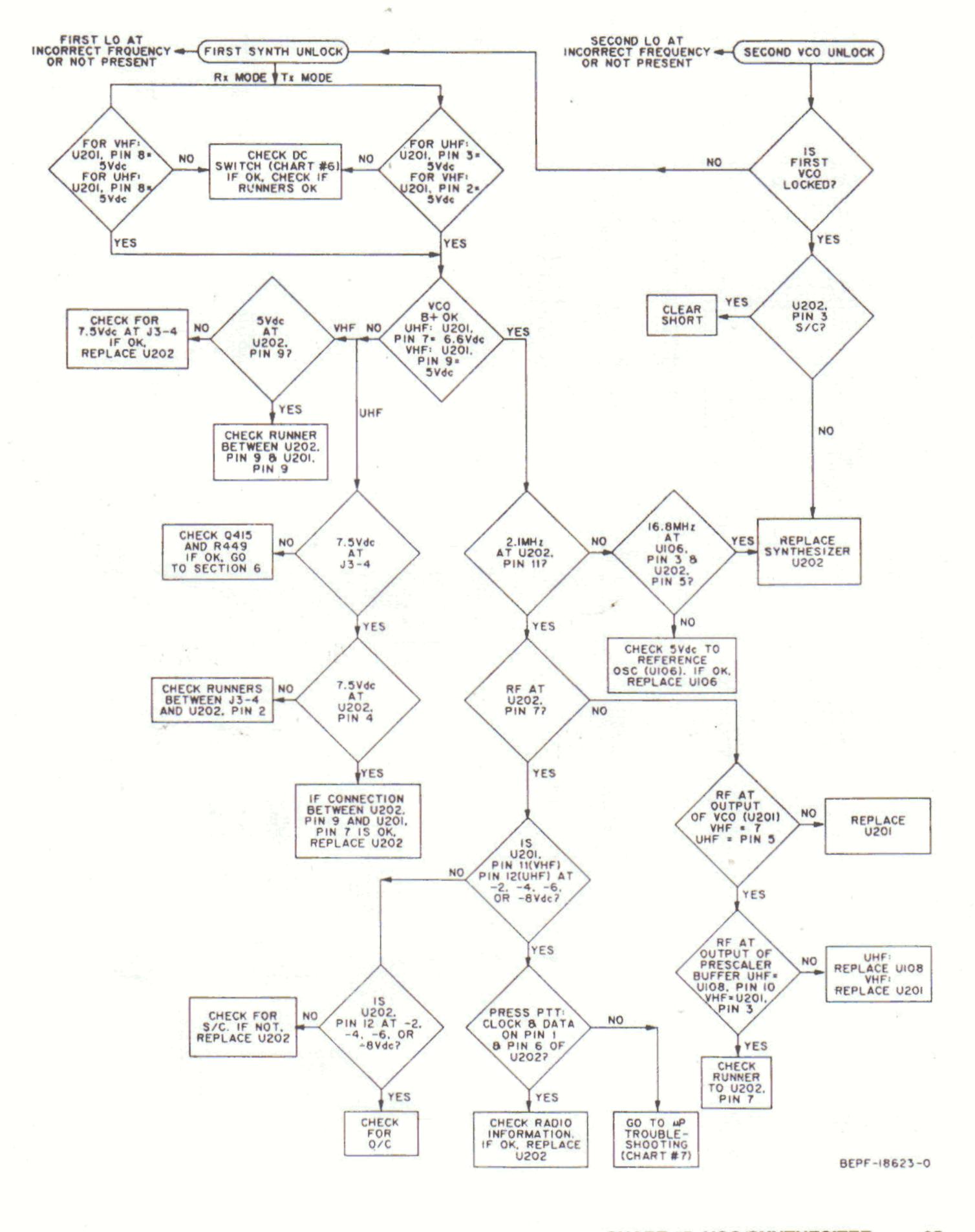


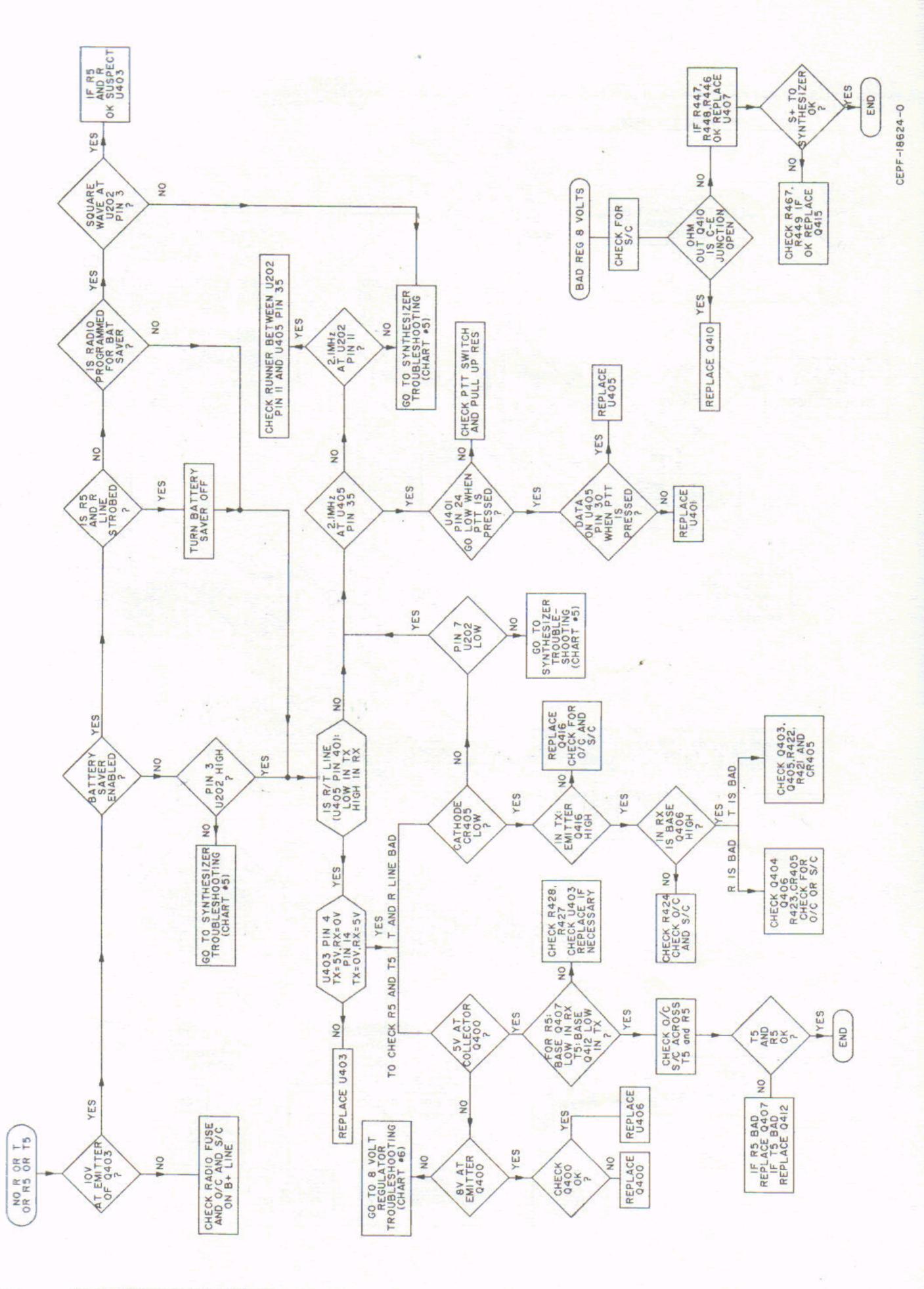


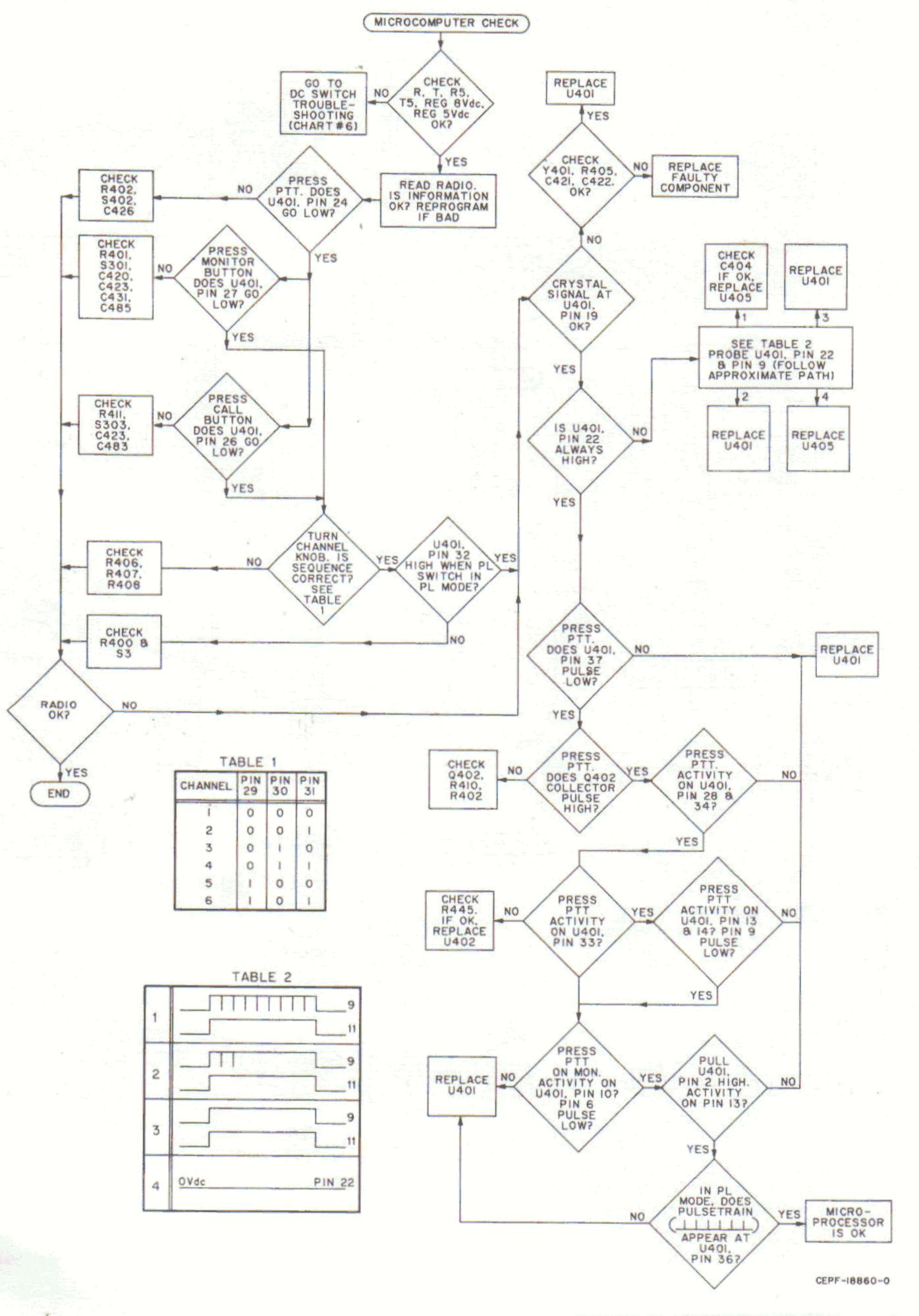


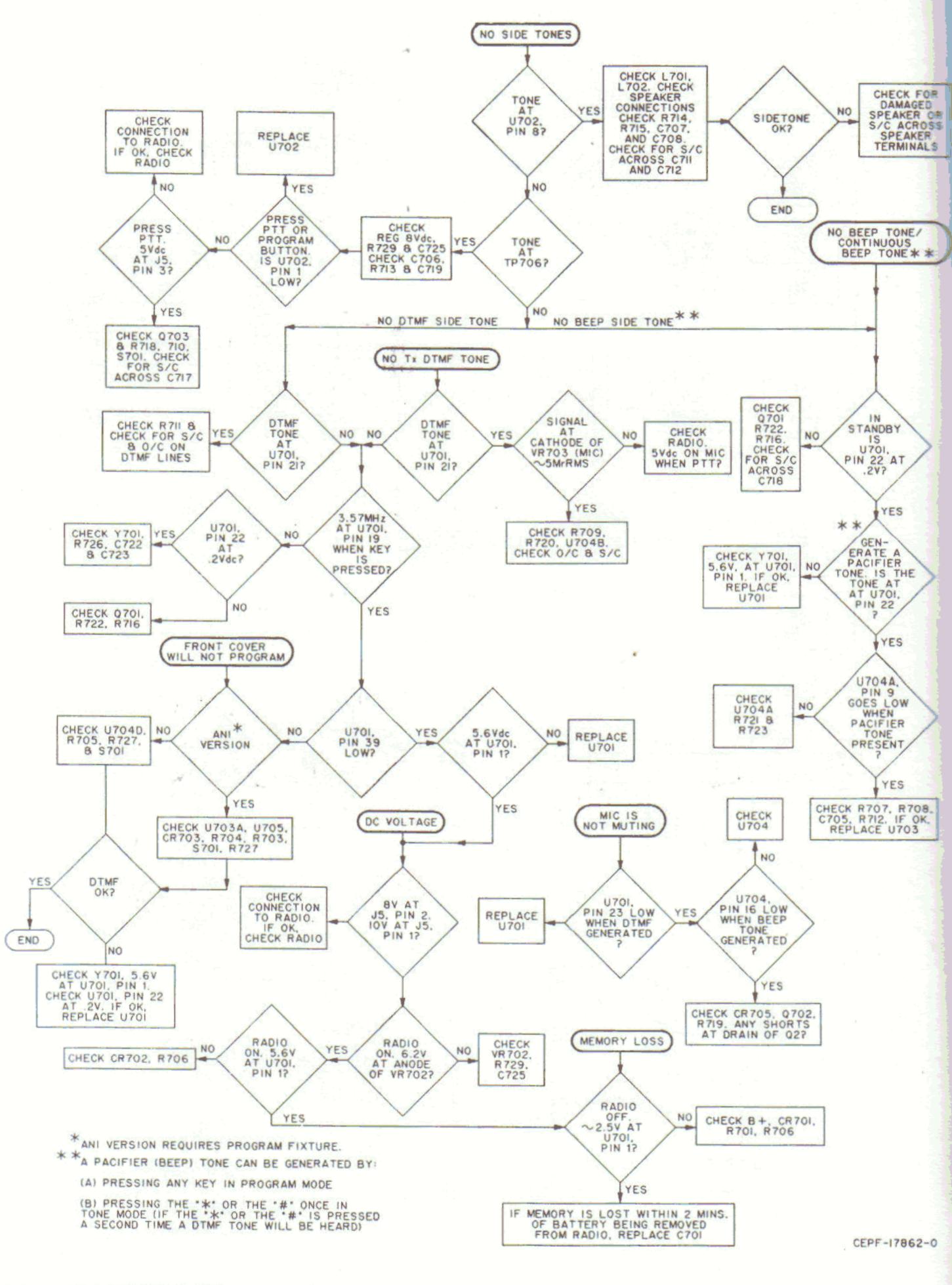


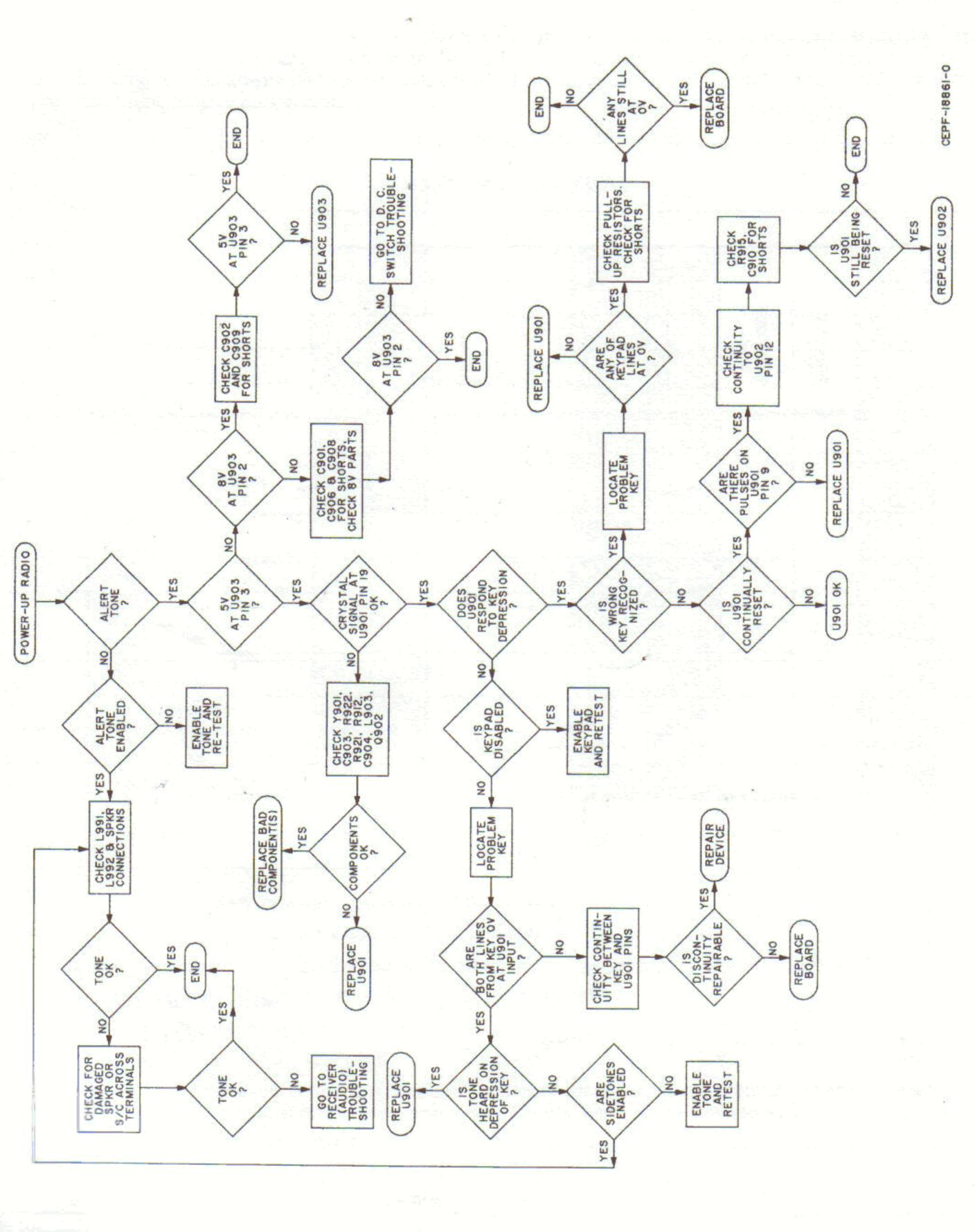












# **NOTES**

# THEORY/MAINTENANCE MANUAL QUESTIONNAIRE

We believe that reports from users provide valuable information for producing quality manuals. By taking a few moments to answer the following questions as they relate to this specific manual, **you** can take an active role in the continuing effort to ensure that our manuals contain the most accurate and complete information of benefit to you. Thank you for your cooperation.

In reference to Manual Number: 68P81046C60-O

# HT800™ Portable Radios

ase check all the propriate boxes:	complete	incomplete	correct	incorrect	8	confusing	size adequate	size too small	not covered
Model/Option Charts	8	jū	8	Ĕ	Clear	8	ad.	ō	où
General Description Features/Options									
Detailed Circuit Descriptions									
Disassembly Procedures				1					
Repair Procedures									
Test Equipment & Service Aids									
Troubleshooting Procedures/Charts									
Illustrations/ Photos/Tables									

2.	Hov	w would you rate	the ov	erall organization	of th	is manual?				
		excellent		very good		good		fair		poor
3.	Is the information in this manual essential to servicing and maintaining the specific equipment?									
		very much so		generally yes		to some extent		no		
4.	Hov	w do you rate this	partic	cular manual?						
		excellent		very good		good		fair		poor
5.	We would appreciate any corrections or recommendations for improving this manual. Please include the specific page number(s) of the procedure, illustration, photo, chart, or table in question.									
	a.	Model/Option Ch	narts:			(Page No				
	b.	General Descrip	tions	(Features/Options	s):	(Page No		)		
	C.	Detailed Circuit	Descr	iptions:		(Page No				
	d.	Disassembly Pro	ocedu	res:	1	(Page No				

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	h.	Illustrations/Photos/Tables:	(Page No.	)	
	g.	Troubleshooting Procedures/C	Charts: (Page No		
	f.	Test Equipment & Service Aid	s: (Page No	)	
	e.	Repair Procedures:	(Page No		
		Please specify the page number	er along with any correct	tions or recommenda	ations for improvement.
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# REPLACEMENT PARTS ORDERING

## - ORDERING INFORMATION —

When ordering replacement parts or equipment information, the complete identification number should be included. This applies to all components, kits, and chassis. If the component part number is not known, the order should include the number of the chassis or kit of which it is a part, and sufficient description of the desired component to identify it.

Crystal and channel element orders should specify the crystal or channel element type number,

crystal and carrier frequency, and the model number in which the part is used.

Orders for active filters, Vibrasender and Vibrasponder resonant reeds should specify type number and frequency, should identify the owner/operator of the communications system in which these items are to be used; and should include any serial numbers stamped on the components being replaced.

# - MAIL ORDERS -

# Send written orders to the following addresses:

Replacement Parts/ Test Equipment/ Crystal Service Items:

Motorola Inc.
Communications Parts Division
Attention: Order Processing
1313 E. Algonquin Road
Schaumburg, IL 60196

Federal Government Orders:

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Communications Parts Division
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Landover, MD 20785

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Communications Parts Division
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# **TELEPHONE ORDERS** -

Replacement Parts/Test Equipment:

1-800-422-4210

1-800-826-1913 (For Federal Government Orders)

Crystal Service Items:

Call: 1-800-323-1570

1-800-445-4564 (For Illinois Residents)

# - TELEX/FAX ORDERS -

Replacement Parts/Test Equipment/ Crystal Service Items:

Telex: 280127 FAX: 312-576-6285

27

Federal Government Orders:

FAX: 301-925-2473 or 301-925-2474

# -CUSTOMER SERVICE

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Crystals:

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